## AK4685

## Multi-channel CODEC with Differential Analog I/O

## GENERAL DESCRIPTION

The AK4685 is a single chip CODEC that integrates 4-channel ADC, 2-channel DAC and a stereo capacitor less headphone amplifier. The converters are designed with an Enhanced Dual Bit architecture for the ADC, and an Advanced Multi-Bit architecture for the DAC's, enabling very low noise performance and achieving wide dynamic range. The differential analog inputs and outputs cancel noise on analog signal lines. Therefore, a stable system can be designed. The AK4685 has a dynamic range of 102dB for ADC, 106 dB for DAC, and is well suited for digital TV and home theater systems.

## FEATURES

$\square$ Asynchronous ADC/DAC1/DAC2 Operation
$\square$ 4ch 24bit ADC

- Differential Input
- 64x Oversampling
- Sampling Rate up to 48kHz
- Linear Phase Digital Anti-Alias Filter
- S/(N+D): 90dB
- Dynamic Range, S/N: 102dB
- Digital HPF for Offset Cancellation
- Channel Independent Digital Volume (+24/-103dB, 0.5dB/step)
- Soft Mute
$\square$ Two 2ch 24bit DAC's
- Differential/Single-end Ouput (DAC1)
- 128x Oversampling
- Sampling Rate up to 192 kHz
- 24bit 8 times Digital Filter
- S/(N+D): 92dB
- Dynamic Range, S/N: 106dB
- Channel Independent Digital Volume (+12/-115dB, $0.5 \mathrm{~dB} /$ step)
- Soft Mute
$\square 40 \mathrm{~mW}$ Capless Stereo Headphone Amplifier
- Output Power: 1.21Vrms @ 3.3V, THD+N(min) = -40dB
- Dynamic Range, S/N: 96dB
- Pop Noise Free at Power-ON/OFF and Mute
$\square$ Independent Mute Pins for 2 lines
$\square$ High Jitter Tolerance
$\square$ TTL Level Digital I/F
$\square$ External Master Clock Input:
256fs, 384fs, 512fs 768fs (fs=32kHz ~ 48kHz)
128fs, 192fs, 256fs 384fs (fs=64kHz ~ 96kHz)
128fs, 192fs (fs=120kHz ~ 192kHz)
$\square$ Audio Serial I/F (PORTA/B/C)
- Master/Slave mode (PORTB)
- I/F format : MSB justified, $I^{2} S$


## $\square I^{2} C$ Bus $\mu P$ I/F for mode setting $\square$ Operating Voltage:

- Digital Out: 3.0V ~5.25V,
- Digital In: 4.75V ~ 5.25V,
- Charge Pump: 3.0V ~3.6V,
- Analog: 4.75V ~ 5.25V

Package: 64pinLQFP (0.5mm pitch)


## ■ Ordering Guide

$$
\begin{array}{lll}
\text { AK4685EQ } & -20 \sim+85^{\circ} \mathrm{C} \quad \text { 64pin LQFP }(0.5 \mathrm{~mm} \text { pitch }) \\
\text { AKD4685 } & \text { Evaluation Board for the AK4685 }
\end{array}
$$

## ■ Pin Layout



## PIN/FUNCTION

| No. | Pin Name | I/O | Function |
| :---: | :---: | :---: | :---: |
| 1 | XTO | O | X'tal Output Pin |
| 2 | MCB/XTI | I | ADC Master Clock Input / X'tal Input Pin |
| 3 | MSB | I | PORTB Master Mode Select Pin. <br> "L" (connected to the ground): Slave mode. <br> "H" (connected to DVDD2): Master mode. |
| 4 | LRCKC | I | DAC2 Input Channel Clock Pin |
| 5 | MCLKC | I | DAC2 Master Clock Input Pin |
| 6 | BICKC | I | DAC2 Audio Serial Data Clock Pin |
| 7 | SDTIC | I | DAC2 Audio Serial Data Input Pin |
| 8 | VSS1 | - | DAC2 Digital Ground Pin, 0V |
| 9 | DVDD3 | - | DAC2 Digital Power Supply Pin, 4.75V~5.25V |
| 10 | SDA | I/O | Control Data Input/Output Pin |
| 11 | SCL | I | Control Data Clock Pin |
| 12 | PDN | I | Power-Down Mode \& Reset Pin <br> When "L", the AK4685 is powered-down, all registers are reset. And then all digital output pins go "L". The AK4685 must be reset once upon power-up. |
| 13 | MT1N | I | ```DAC1 Mute Pin "H": Normal Operation "L": Mute``` |
| 14 | MT2N | I | $\begin{aligned} & \text { DAC2 Mute Pin } \\ & \text { "H": Normal Operation } \\ & \text { "L": Mute } \end{aligned}$ |
| 15 | DVDD2 | - | DAC1 Digital Power Supply Pin, 4.75V~5.25V |
| 16 | VSS2 | - | DAC1 Digital Ground Pin, 0V |
| 17 | SDTIA | I | DAC1 Audio Serial Data Input Pin |
| 18 | BICKA | I | DAC1 Audio Serial Data Clock Pin |
| 19 | MCLKA | I | DAC1 Master Clock Input Pin |
| 20 | LRCKA | I | DAC1 Input Channel Clock Pin |
| 21 | SGL | I | Analog Output Mode Select Pin. <br> "L" (connected to the ground): Differential mode. <br> "H" (connected to DVDD): Single End mode. |
| 22 | AVDD2 | - | DAC1 Analog Power Supply Pin, 4.75V $\sim 5.25 \mathrm{~V}$ |
| 23 | VSS3 | - | DAC1 Analog Ground Pin, 0V |
| 24 | ROUT+ | O | Rch Positive Analog Output Pin |
| 25 | ROUT- | O | Rch Negative Analog Output Pin |
| 26 | LOUT+ | O | Lch Positive Analog Output Pin |
| 27 | LOUT- | O | Lch Negative Analog Output Pin |
| 28 | NC | - | No internal bonding. This pin must be connected to Ground. |
| 29 | CP | I | Positive Charge Pump Capacitor Terminal Pin |
| 30 | CN | I | Negative Charge Pump Capacitor Terminal Pin |
| 31 | PVDD | - | Charge Pump Power Supply Pin, 3.0V ~3.6V. |
| 32 | VSS4 | - | Charge Pump Ground Pin, 0V. |
| 33 | PVEE | - | Charge Pump Negative Power Output Pin. |
| 34 | HPL | O | Lch Headphone-Amp Output Pin |
| 35 | HPR | O | Rch Headphone-Amp Output Pin |
| 36 | AVDD3 | - | DAC2 Analog Power Supply Pin, 4.75V $\sim 5.25 \mathrm{~V}$ |
| 37 | VSS5 | - | DAC2 Analog Ground Pin, 0V |
| 38 | RVCOM2 | O | Rch VCOM Output 2 Pin |
| 39 | RIN2+ | 1 | Rch Positive Analog Input 2 Pin |
| 40 | RIN2- | I | Rch Negative Analog Input 2 Pin |
| 41 | RAMPO2 | O | Rch Pre-Amp Output 2 Pin |
| 42 | LAMPO2 | O | Lch Pre-Amp Output 2 Pin |


| 43 | LIN2- | I | Lch Negative Analog Input 2 Pin |
| :--- | :--- | :---: | :--- |
| 44 | LIN2 + | I | Lch Positive Analog Input 2 Pin |
| 45 | LVCOM2 | O | Lch VCOM Output 2 Pin |
| 46 | VCOM | - | DAC/ADC Common Voltage Output Pin. AVDD1 x 0.5(typ). <br> $10 \mu$ F capacitor should be connected to VSS6 externally. |
| 47 | VSS6 | - | ADC Analog Ground Pin, 0V |
| 48 | AVDD1 | - | ADC Analog Power Supply Pin, 4.75V~5.25V |
| 49 | RVCOM1 | O | Rch VCOM Output 1 Pin |
| 50 | RIN1+ | I | Rch Positive Analog Input 1 Pin |
| 51 | RIN1- | I | Rch Negative Analog Input 1 Pin |
| 52 | RAMPO1 | O | Rch Pre-Amp Output 1 Pin |
| 53 | LAMPO1 | O | Lch Pre-Amp Output 1 Pin |
| 54 | LIN1- | I | Lch Negative Analog Input 1 Pin |
| 55 | LIN1+ | I | Lch Positive Analog Input 1 Pin |
| 56 | LVCOM1 | O | Lch VCOM Output 1 Pin |
| 57 | LRCKB | I/O | ADC Channel Clock Pin |
| 58 | BICKB | I/O | ADC Audio Serial Data Clock Pin |
| 59 | SDTOB2 | O | ADC Audio Serial Data Output 2 Pin |
| 60 | SDTOB1 | O | ADC Audio Serial Data Output 1 Pin |
| 61 | MCKO | O | Master Clock Output Pin |
| 62 | TVDD | - | Output Buffer Power Supply Pin, 3.0V~5.25V |
| 63 | VSS7 | - | ADC Digital Ground Pin, 0V |
| 64 | DVDD1 | - | ADC Digital Power Supply Pin, 4.75V~5.25V |

Note 1. All digital input pins must not be left floating.
Note 2. AC coupling capacitors should be connected to analog input pins (LIN1+/-, LIN $2+/-$, RIN1+/-, RIN $2+/-$ ).
Note 3. AC coupling capacitors should be connected to analog output pins (LOUT $+/-$, ROUT $+/-$ ).

## ■ Handling of Unused Pin

The unused I/O pins must be processed appropriately as below.

| Classification | Pin Name | Setting |
| :--- | :--- | :--- |
| Analog | LOUT+/-, ROUT+/-, | These pins should be open. |
|  | LIN1+, LIN 2+, RIN1+, RIN 2+ | These pins should be connected to each <br> VCOMO pins (LVCOM1/2,RVCOM1/2) |
|  | LIN1-, LIN 2-, RIN1-, RIN 2- | These pins should be connected to each <br> AMPO pins (LAMPO1/2,RAMPO1/2) |
|  | SDTOB1, SDTOB2, XTO, MCLKO, <br> LRCKB(Master), BICKB(Master) | These pins should be open. |
| MCLKA/C, MCB, LRCKA-C(Slave), <br> BICKA-C(Slave), SDTIA,C, MSB, MT1N, MT2N, <br> SGL | These pins should be connected to ground. |  |
|  | SDA, SCL | These pins should be pulled-up to DVDD3. |
| - | NC | This pin should be connected to ground. |

## ■ Power-down Pin States

| No. | Pin Name | I/O | Power-down (PDN pin = "L") |
| :---: | :---: | :---: | :---: |
| 1 | XTO | O | H (DVDD1) |
| 2 | MCB/XTI | I | Pull-down 25kS(typ) to VSS7 |
| 3 | MSB | I | Hi-Z |
| 4 | LRCKC | I | Hi-Z |
| 5 | MCLKC | I | Hi-Z |
| 6 | BICKC | I | Hi-Z |
| 7 | SDTIC | 1 | Hi-Z |
| 8 | VSS1 | - | - |
| 9 | DVDD3 | - | - |
| 10 | SDA | I/O | Hi-Z |
| 11 | SCL | I | Hi-Z |
| 12 | PDN | I | - |
| 13 | MT1N | I | Hi-Z |
| 14 | MT2N | I | Hi-Z |
| 15 | DVDD2 | - | - |
| 16 | VSS2 | - | - |
| 17 | SDTIA | I | Hi-Z |
| 18 | BICKA | I | Hi-Z |
| 19 | MCLKA | I | Hi-Z |
| 20 | LRCKA | I | Hi-Z |
| 21 | SGL | I | Hi-Z |
| 22 | AVDD2 | - | - |
| 23 | VSS3 | - | - |
| 24 | LOUT+ | O | pull up 190k $\Omega$ (typ) to AVDD2 |
| 25 | LOUT- | O | Hi-Z |
| 26 | ROUT+ | O | pull up 190k 2 (typ) to AVDD2 |
| 27 | ROUT- | O | Hi-Z |
| 28 | NC | - | Hi-Z |
| 29 | CP | I | Pull-up $80 \Omega$ (typ) to PVDD |
|  | CN | I | Pull-down $80 \Omega$ (typ) to VSS4 |
| 31 | PVDD | - | - |
| 32 | VSS4 | - | - |
| 33 | PVEE | O | Pull-down 17.5kS(typ) to VSS4 |
| 34 | HPL | O | Pull-down 20ת(typ) to VSS5 |
| 35 | HPR | O | Pull-down $20 \Omega$ (typ) to VSS5 |
| 36 | AVDD3 | - | - |
| 37 | VSS5 | - | - |
| 38 | RVCOM2 | O | Hi-Z |
| 39 | RIN2+ | I | Hi-Z |
| 40 | RIN2- | I | Hi-Z |
| 41 | RAMPO2 | O | Hi-Z |
| 42 | LAMPO2 | O | Hi-Z |
| 43 | LIN2- | I | Hi-Z |
| 44 | LIN2+ | I | Hi-Z |
| 45 | LVCOM2 | O | Hi-Z |
| 46 | VCOM | - | Pull-down 500 (typ) to VSS6 |
| 47 | VSS6 | - | - |
| 48 | AVDD1 | - | - |
| 49 | RVCOM1 | O | Hi-Z |
| 50 | RIN1+ | I | Hi-Z |


| No. | Pin Name | I/O | Power-down (PDN pin = "L") |
| :---: | :---: | :---: | :---: |
| 51 | RIN1- | I | Hi-Z |
| 52 | RAMPO1 | O | Hi-Z |
| 53 | LAMPO1 | O | Hi-Z |
| 54 | LIN1- | I | Hi-Z |
| 55 | LIN1+ | I | Hi-Z |
| 56 | LVCOM1 | O | Hi-Z |
| 57 | LRCKB | I/O | $\begin{gathered} \hline \text { Hi-Z (MSB pin }=\text { "L") } \\ \text { L (MSB pin }=\text { "H") } \\ \hline \end{gathered}$ |
| 58 | BICKB | I/O | $\begin{gathered} \hline \text { Hi-Z (MSB pin }=" L ") \\ \text { L (MSB pin }=\text { "H") } \\ \hline \end{gathered}$ |
| 59 | SDTOB2 | O | L |
| 60 | SDTOB1 | O | L |
| 61 | MCKO | O | L (X'tal mode) MCB through (External CLK mode) |
| 62 | TVDD | - | - |
| 63 | VSS7 | - | - |
| 64 | DVDD1 | - | - |

Note 1. All digital input pins must not be left floating.
Note 4. The differential output pins of analog line-out (LOUT $+\leftrightarrow$ LOUT- and ROUT $+\leftrightarrow$ ROUT-) are connected internally via $150 \mathrm{k} \Omega$ (typ) resistors.
Note 5. Preamplifier output pins (LAMPO1 $\leftrightarrow$ LVCOM1, RAMPO1 $\leftrightarrow$ RVCOM1, LAMPO2 $\leftrightarrow$ LVCOM2, RAMPO2 $\leftrightarrow$ RVCOM 2 ) are connected internally via $200 \mathrm{k} \Omega$ (typ) resistors.

## ABSOLUTE MAXIMUM RATINGS

(VSS1-7=0V; Note 6)

| Parameter | Symbol | min | max | Units |
| :--- | :---: | :---: | :---: | :---: |
| Power Supply | TVDD | -0.3 | 6.0 | V |
|  | DVDD1 | -0.3 | 6.0 | V |
|  | DVDD2 | -0.3 | 6.0 | V |
|  | DVDD3 | -0.3 | 6.0 | V |
|  | AVDD1 | -0.3 | 6.0 | V |
|  | AVDD2 | -0.3 | 6.0 | V |
| AVDD3 | -0.3 | 6.0 | V |  |
| Input Current (any pins except for supplies) | PVDD | -0.3 | 4.0 | V |
| Digital Input Voltage 1 <br> (MCB/XTI, MSB pins) | VIND1 | -0.3 | DVDD1+0.3 | V |
| Digital Input Voltage 2 <br> (SDTIA, BICKA, MCLKA, LRCKA and SGL pins) | VIND2 | -0.3 | DVDD2+0.3 | V |
| Digital Input Voltage 3 <br> (LRCKC, MCLKC, BICKC, SDTIC, SDA, SCL, PDN, MT1N <br> and MT2N pins) | VIND3 | -0.3 | DVDD3+0.3 | V |
| Digital Input Voltage 4 <br> (LRCKB, BICKB pins) | VIND4 | -0.3 | TVDD+0.3 | V |
| Analog Input Voltage <br> (LIN1+/1-/2+/2-, RIN1+/1-/2+/2- pins) | VINA1 | -0.3 | AVDD1+0.3 | V |
| Ambient Operating Temperature | Ta | -20 | 85 | ${ }^{\circ} \mathrm{C}$ |
| Storage Temperature | Tstg | -65 | 150 | ${ }^{\circ} \mathrm{C}$ |

Note 6. VSS1-7 must be connected to the same analog ground plane.
WARNING: Operation at or beyond these limits may result in permanent damage to the device.
Normal operation is not guaranteed at these extremes.

## RECOMMENDED OPERATING CONDITIONS

(VSS1-7=0V; Note 6)

| Parameter | Symbol | min | typ | max | Units |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Power Supply (Table 4, Note 8) | TVDD | 3.0 | 3.3 | 5.25 | V |
|  | DVDD1 | 4.75 | 5.0 | 5.25 | V |
|  | DVDD2 | 4.75 | 5.0 | 5.25 | V |
|  | DVDD3 | 4.75 | 5.0 | 5.25 | V |
|  | AVDD1 | 4.75 | 5.0 | 5.25 | V |
|  | AVDD2 | 4.75 | 5.0 | 5.25 | V |
|  | AVDD3 | 4.75 | 5.0 | 5.25 | V |
|  | PVDD | 3.0 | 3.3 | 3.6 | V |

Note 7. The AVDD1, AVDD2, AVDD3, DVDD1, DVDD2 and DVDD3 must be the same voltage.
The TVDD must not exceed any of AVDD1, AVDD2, AVDD3, DVDD1, DVDD2 and DVDD3 voltage.
Note 8. The power-up sequences of AVDD1-3, DVDD1-3, PVDD and TVDD are not important. All power supply pins must be up when the PDN pin= "L".
*AKM assumes no responsibility for the usage beyond the conditions in this datasheet.

## ANALOG CHARACTERISTICS

(Ta $=25^{\circ} \mathrm{C} ; \mathrm{TVDD}=3.3 \mathrm{~V} ;$ DVDD1-3 $=\mathrm{AVDD} 1-3=5.0 \mathrm{~V} ; \mathrm{PVDD}=3.3 \mathrm{~V} ; \mathrm{VSS} 1-7=0 \mathrm{~V} ; \mathrm{fs}=48 \mathrm{kHz} ; \mathrm{BICK}=64 \mathrm{fs} ;$ Signal Frequency $=1 \mathrm{kHz}$; 24bit Data; Measurement Frequency $=20 \mathrm{~Hz} \sim 20 \mathrm{kHz}$ at $\mathrm{fs}=48 \mathrm{kHz}, 20 \mathrm{~Hz} \sim 40 \mathrm{kHz}$ at $\mathrm{fs}=96 \mathrm{kHz}$; $20 \mathrm{~Hz} \sim 40 \mathrm{kHz}$ at $\mathrm{fs}=192 \mathrm{kHz}$, all blocks are synchronized, unless otherwise specified)

| Parameter |  | min | typ | max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ADC Characteristics |  |  |  |  |  |
| Feedback Resistance |  | 10 |  | 50 | $\mathrm{k} \Omega$ |
| Resolution |  |  |  | 24 | Bits |
| S/(N+D) (-1dBFS. Note 9) | $\mathrm{fs}=48 \mathrm{kHz}$ | 82 | 95 |  | dB |
| DR (-60dBFS. Note 9) | $\mathrm{fs}=48 \mathrm{kHz}$, A-weighted | 93 | 102 |  | dB |
| S/N (input off) | $\mathrm{fs}=48 \mathrm{kHz}$, A-weighted | 93 | 102 |  | dB |
| Interchannel Isolation | (Note 10) | 90 | 100 |  | dB |
| Interchannel Gain Mismatch |  |  |  | 0.6 | dB |
| Gain Drift |  |  | 50 | - | $\mathrm{ppm} /{ }^{\circ} \mathrm{C}$ |
| Input Range (Pre- | $($ Pre-Amp output) $= \pm 3.3 \times$ AVDD $1 / 5$ | $\pm 2.97$ | $\pm 3.3$ | $\pm 3.63$ | Vpp |
| Power Supply Rejection | (Note 12) |  | 60 |  | dB |
| DAC to Analog Output Characteristics (Differential Mode) |  |  |  |  |  |
| Resolution |  |  |  | 24 | Bits |
| $\mathrm{S} /(\mathrm{N}+\mathrm{D}) \quad$ (0dBFS) | $\begin{aligned} & \mathrm{fs}=48 \mathrm{kHz} \\ & \mathrm{fs}=96 \mathrm{kHz} \\ & \mathrm{fs}=192 \mathrm{kHz} \end{aligned}$ | $82$ | $\begin{aligned} & \hline 95 \\ & 95 \\ & 95 \\ & \hline \end{aligned}$ |  | $\begin{aligned} & \hline \mathrm{dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \\ & \hline \end{aligned}$ |
| DR (-60dBFS) | $\begin{aligned} & \mathrm{fs}=48 \mathrm{kHz}, \mathrm{~A} \text {-weighted } \\ & \mathrm{fs}=96 \mathrm{kHz} \\ & \mathrm{fs}=96 \mathrm{kHz}, \text { A-weighted } \\ & \mathrm{fs}=192 \mathrm{kHz} \\ & \mathrm{fs}=192 \mathrm{kHz} \text {, A-weighted } \end{aligned}$ | $98$ | $\begin{aligned} & 107 \\ & 102 \\ & 107 \\ & 102 \\ & 107 \\ & \hline \end{aligned}$ |  | $\begin{aligned} & \mathrm{dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \\ & \hline \end{aligned}$ |
| S/N ("0" data) | $\begin{aligned} & \mathrm{fs}=48 \mathrm{kHz}, \mathrm{~A} \text {-weighted } \\ & \mathrm{fs}=96 \mathrm{kHz} \\ & \mathrm{fs}=96 \mathrm{kHz}, \mathrm{~A} \text {-weighted } \\ & \mathrm{fs}=192 \mathrm{kHz} \\ & \mathrm{fs}=192 \mathrm{kHz} \text {, A-weighted } \end{aligned}$ | $98$ | $\begin{aligned} & 108 \\ & 102 \\ & 107 \\ & 102 \\ & 107 \\ & \hline \end{aligned}$ |  | $\begin{aligned} & \mathrm{dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \end{aligned}$ |
| Interchannel Isolation |  | 90 | 100 |  | dB |
| Interchannel Gain Mismatch |  |  |  | 0.5 | dB |
| Gain Drift |  |  | 50 | - | $\mathrm{ppm} /{ }^{\circ} \mathrm{C}$ |
| Output Voltage | AOUT-) $= \pm 2.56 \times$ AVDD2/5 | $\pm 2.30$ | $\pm 2.56$ | $\pm 2.82$ | Vpp |
| Load Resistance | (AC Load, Note 11) | 5 |  |  | $\mathrm{k} \Omega$ |
| Load Capacitance |  |  |  | 30 | pF |
| Power Supply Rejection | (Note 12) |  | 50 |  | dB |
| DAC to Analog Output Characteristics (Single End Mode) |  |  |  |  |  |
| Resolution |  |  |  | 24 | Bits |
| $\mathrm{S} /(\mathrm{N}+\mathrm{D}) \quad$ (0dBFS) | $\begin{aligned} & \mathrm{fs}=48 \mathrm{kHz} \\ & \mathrm{fs}=96 \mathrm{kHz} \\ & \mathrm{fs}=192 \mathrm{kHz} \end{aligned}$ | $80$ | $\begin{aligned} & \hline 90 \\ & 90 \\ & 90 \\ & \hline \end{aligned}$ |  | $\begin{aligned} & \hline \mathrm{dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \end{aligned}$ |
| DR (-60dBFS) | $\begin{aligned} & \text { fs }=48 \mathrm{kHz}, \mathrm{~A} \text {-weighted } \\ & \mathrm{fs}=96 \mathrm{kHz} \\ & \mathrm{fs}=96 \mathrm{kHz} \text {, A-weighted } \\ & \mathrm{fs}=192 \mathrm{kHz} \\ & \mathrm{fs}=192 \mathrm{kHz} \text {, A-weighted } \end{aligned}$ | $98$ | $\begin{gathered} \hline 105 \\ 99 \\ 105 \\ 99 \\ 105 \\ \hline \end{gathered}$ |  | $\begin{aligned} & \mathrm{dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \\ & \hline \end{aligned}$ |
| S/N ("0" data) | $\begin{aligned} & \text { fs }=48 \mathrm{kHz}, \mathrm{~A} \text {-weighted } \\ & \mathrm{fs}=96 \mathrm{kHz} \\ & \mathrm{fs}=96 \mathrm{kHz} \text {, A-weighted } \\ & \mathrm{fs}=192 \mathrm{kHz} \\ & \mathrm{fs}=192 \mathrm{kHz} \text {, A-weighted } \end{aligned}$ | $98$ | $\begin{gathered} \hline 106 \\ 99 \\ 105 \\ 99 \\ 105 \\ \hline \end{gathered}$ |  | $\begin{aligned} & \mathrm{dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \\ & \hline \end{aligned}$ |
| Interchannel Isolation |  | 90 | 100 |  | dB |
| Interchannel Gain Mismatch |  |  |  | 0.5 | dB |


| Gain Drift |  | 50 | - | $\mathrm{ppm} /{ }^{\circ} \mathrm{C}$ |
| :---: | :---: | :---: | :---: | :---: |
| Output Voltage ${ }^{\text {AOUT }}+=2.83 \times$ AVDD2/5 | 2.54 | 2.83 | 3.12 | Vpp |
| Load Resistance (AC Load. Note 11) | 5 |  |  | $\mathrm{k} \Omega$ |
| Load Capacitance |  |  | 30 | pF |
| Power Supply Rejection (Note 12) |  | 50 |  | dB |
| DAC to Headphone Output (HPL, HPR pin) Characteristics (Note 13) |  |  |  |  |
| $\mathrm{S} /(\mathrm{N}+\mathrm{D})$ (0dBFS. Note 14) <br> $(-6 d B F S . ~ N o t e ~ 14)$ | $\begin{aligned} & \hline 40 \\ & 60 \end{aligned}$ | $\begin{aligned} & \hline 60 \\ & 66 \end{aligned}$ |  | $\begin{aligned} & \hline \mathrm{dB} \\ & \mathrm{~dB} \end{aligned}$ |
| S/N ("0" data, A-weighted) | 88 | 98 |  | dB |
| Interchannel Isolation | 60 | 80 |  | dB |
| Interchannel Gain Mismatch |  |  | 0.8 | dB |
| Output Voltage $\quad$ AOUT=1.21 x AVDD3/5 (Note 15) |  | 1.21 |  | Vrms |
| Load Resistance (Note 16) | 32 |  |  | $\Omega$ |
| Load Capacitance |  |  | 300 | pF |
| Power Supply Rejection (Note 12) |  | 50 |  | dB |

Note 9 . When 33 kHz is input to the external input resistor (Ri), 36 kHz is input to the feed back resistor (Rf) and $+/-2.70 \mathrm{Vpp}(-1 \mathrm{~dB})$ or $+/-0.003 \mathrm{Vpp}(-60 \mathrm{~dB})$ is input to the differential input ports.


Figure 1. ADC Input Circuit
Note 10. This value is the inter-channel isolation between all the channels of the LIN1-2 and RIN1-2.
Note 11. Load resistance via an AC coupling resistor.
Note 12. PSR is applied to AVDD1, AVDD2, AVDD3, DVDD1, DVDD2, DVDD3 and PVDD with $1 \mathrm{kHz}, 50 \mathrm{mVpp}$.
Note $13.6 .8 \Omega$ resistors should be connected in direct to the headphone output pins. When $\mathrm{fs}=48 \mathrm{kHz}, 96 \mathrm{kHz}$ or 192 kHz , the measurement frequency of headphone output is $20 \mathrm{~Hz} \sim 20 \mathrm{kHz}$.
Note 14. When Load Resistance $=6.8 \Omega+32 \Omega$.
Note $15.1 .21 \mathrm{Vrms}(\operatorname{typ})$ is output to the output pin. When load resistance $=6.8 \Omega+32 \Omega, 1 \mathrm{Vrms}(\operatorname{typ})$ is output at $32 \Omega$ output port.
Note 16. A more than $32 \Omega$ device can be connected after a $6.8 \Omega$ resistor in direct.

| Power Supplies |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Parameter | min | typ | max | Units |
| Power Supply Current |  |  |  |  |
| Normal Operation (PDN pin = "H") |  |  |  |  |
| TVDD (Note 17) |  | 6 | 9 | mA |
| DVDD1+AVDD1 |  | 33 | 47 | mA |
| DVDD2+AVDD2 |  |  |  |  |
| Differential mode $\mathrm{fs}=48 \mathrm{kHz}$ |  | 13 | 19 | mA |
| $\mathrm{fs}=96 \mathrm{kHz}$ |  | 14 | - | mA |
| $\mathrm{fs}=192 \mathrm{kHz}$ |  | 17 | - | mA |
| Single End mode fs $=48 \mathrm{kHz}$ |  | 14 | - | mA |
| $\mathrm{fs}=96 \mathrm{kHz}$ |  | 15 | - | mA |
| $\mathrm{fs}=192 \mathrm{kHz}$ |  | 18 | - | mA |
| DVDD3+AVDD3 (No Input) |  |  |  |  |
| $\mathrm{fs}=48 \mathrm{kHz}$ |  | 20 | 30 | mA |
| $\mathrm{fs}=96 \mathrm{kHz}$ |  | 21 | - | mA |
| $\mathrm{fs}=192 \mathrm{kHz}$ |  | 23 | - | mA |
| PVDD |  | 6 | 9 | mA |
| Power-Down Mode (PDN pin = "L"; Note 18) |  |  |  |  |
| TVDD |  | 10 | 100 | $\mu \mathrm{A}$ |
| DVDD1+AVDD1 |  | 10 | 100 | $\mu \mathrm{A}$ |
| DVDD2+AVDD2 |  | 10 | 100 | $\mu \mathrm{A}$ |
| DVDD3+AVDD3 |  | 10 | 100 | $\mu \mathrm{A}$ |
| PVDD |  | 10 | 100 | $\mu \mathrm{A}$ |

Note 17. Master Mode. MCB=36.864MHz. 20pF load capacitors are connected to MCKO, BICKB, LRCKB, SDTOB1 and SDTOB2 pins.
Note 18. All digital inputs including clock pins (MCLKA, MCB, MCLKC, BICKA, BICKB, BICKC, LRCKA, LRCKB, LRCKC, SDTIA, SDTIC) are held at DVDD1, DVDD2, DVDD3, VSS1, VSS2 or VSS7.

## FILTER CHARACTERISTICS

( $\mathrm{Ta}=-20^{\circ} \mathrm{C} \sim+85^{\circ} \mathrm{C}$; TVDD $=3.0 \sim 5.25 \mathrm{~V}$; DVDD1-3=AVDD1-3=4.75~5.25V; PVDD=3.0~3.6V; fs $=48 \mathrm{kHz}$ )

| Parameter | Symbol | min | typ | max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ADC Digital Filter (Decimation LPF): |  |  |  |  |  |
| Passband (Note 19) $\pm 0.1 \mathrm{~dB}$ <br>   -0.2 dB <br>  -3.0 dB  | PB | $0$ | $\begin{array}{r} 20.0 \\ 23.0 \\ \hline \end{array}$ | $\overline{18.9}$ | $\begin{aligned} & \mathrm{kHz} \\ & \mathrm{kHz} \\ & \mathrm{kHz} \end{aligned}$ |
| Stopband | SB | 28.0 |  |  | kHz |
| Stopband Attenuation | SA | 68 |  |  | dB |
| Group Delay (Note 20) | GD |  | 16 |  | 1/fs |
| ADC Digital Filter (HPF): |  |  |  |  |  |
| Frequency Response (Note 19) -3 dB <br> -0.1 dB  | FR |  | $\begin{aligned} & \hline 1.0 \\ & 6.5 \\ & \hline \end{aligned}$ |  | $\begin{aligned} & \mathrm{Hz} \\ & \mathrm{~Hz} \\ & \hline \end{aligned}$ |
| DAC Digital Filter: |  |  |  |  |  |
| Passband (Note 19) $\pm 0.1 \mathrm{~dB}$ <br>   | PB | $0$ | 24.0 | $21.8$ | $\begin{aligned} & \hline \mathrm{kHz} \\ & \mathrm{kHz} \\ & \hline \end{aligned}$ |
| Stopband | SB | 26.2 |  |  | kHz |
| Stopband Attenuation | SA | 54 |  |  | dB |
| Group Delay (Note 20) | GD |  | 20 |  | 1/fs |
| DAC Digital Filter + Analog Filter: |  |  |  |  |  |
| Frequency Response: $0 \sim 20.0 \mathrm{kHz}$ <br>  $40.0 \mathrm{kHz} \quad$ (Note 21) <br>  80.0 kHz (Note 21) | $\begin{aligned} & \hline \text { FR } \\ & \text { FR } \\ & \text { FR } \end{aligned}$ |  | $\begin{aligned} & \pm 0.2 \\ & \pm 0.3 \\ & \pm 1.0 \end{aligned}$ |  | $\begin{aligned} & \mathrm{dB} \\ & \mathrm{~dB} \\ & \mathrm{~dB} \end{aligned}$ |

Note 19. The passband and stopband frequencies scale with fs.
For example, 21.8 kHz at -0.1 dB is 0.454 x fs (DAC). The reference frequency of these responses is 1 kHz .

Note 20. The calculating delay time occurred at digital filtering. This time is from setting the input of analog signal to setting the 24bit data of both channels to the output register of PORTB.
For DAC, this time is from setting the 20/24bit data of both channels on input registers of PORTA and PORTC to the output of analog signal.
Note 21.40.0kHz@fs=96kHz, 80.0kHz@fs=192kHz.

## DC CHARACTERISTICS

( $\mathrm{Ta}=-20^{\circ} \mathrm{C} \sim+85^{\circ} \mathrm{C}$; TVDD $=3.0 \sim 5.25 \mathrm{~V}$; DVDD1-3 $=4.75 \sim 5.25 \mathrm{~V}$, AVDD1-3=4.75~5.25V; PVDD=3.0~3.6V)

| Parameter | Symbol | min | typ | max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: |
| High-Level Input Voltage (Except XTI pin) | VIH | 2.2 | - | - | V |
| (XTI pin) | VIH | 70\%DVDD1 | - | - | V |
| Low-Level Input Voltage (Except XTI pin) | VIL | - | - | 0.8 | V |
| (XTI pin) | VIL | - | - | 30\%DVDD1 | V |
| Input Voltage at AC Coupling (XTI pin) (Note 22) | VAC | 40\%DVDD1 | - | - | Vpp |
| High-Level Output Voltage (Iout $=-400 \mu \mathrm{~A}$. Except XTO pin) | VOH | TVDD-0.4 | - | - | V |
| $\qquad$ | VOL | - | - | 0.4 | V |
| Input Leakage Current | Iin | - | - | $\pm 10$ | $\mu \mathrm{A}$ |

Note 22 . This is the value when a capacitor $(0.1 \mu \mathrm{~F})$ is connected to the XTI pin.

## SWITCHING CHARACTERISTICS

(Ta $=-20^{\circ} \mathrm{C} \sim+85^{\circ} \mathrm{C} ;$ TVDD $=3.0 \sim 5.25 \mathrm{~V} ; \mathrm{DVDD} 1-3=\mathrm{AVDD} 1-3=4.75 \sim 5.25 \mathrm{~V} ; \mathrm{PVDD}=3.0 \sim 3.6 \mathrm{~V} ; \mathrm{C}_{\mathrm{L}}=20 \mathrm{pF}$ (except for SDA pin), $\mathrm{Cb}=400 \mathrm{pF}($ SDA pin))

| Parameter | Symbol | min | typ | max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Master Clock Timing |  |  |  |  |  |
| Crystal Resonator $\quad$ Frequency | fXTAL | 11.2896 |  | 24.576 | MHz |
| External Clock $\quad \begin{aligned} & \text { Frequency } \\ & \text { Duty }\end{aligned}$ | fECLK dECLK | $\begin{gathered} 8.192 \\ 40 \end{gathered}$ | 50 | $\begin{gathered} 36.864 \\ 60 \\ \hline \end{gathered}$ | $\begin{gathered} \mathrm{MHz} \\ \% \end{gathered}$ |
| MCKO Output( $\begin{aligned} & \text { Frequency } \\ & \text { Duty }\end{aligned}$ | fMCK dMCLK | $\begin{gathered} \hline 8.192 \\ 40 \\ \hline \end{gathered}$ | 50 | $\begin{gathered} \hline 36.864 \\ 60 \\ \hline \end{gathered}$ | $\begin{gathered} \mathrm{MHz} \\ \% \\ \hline \end{gathered}$ |
| Master Clock (Note 23) 256fsn, 128fsd: <br> Pulse Width Low <br> Pulse Width High 384fsn, 192fsd: <br> Pulse Width Low Pulse Width High 512fsn, 256fsd, 128fsq: Pulse Width Low Pulse Width High 768fsn, 384fsd, 192fsq: Pulse Width Low Pulse Width High | fCLK tCLKL tCLKH fCLK tCLKL tCLKH fCLK tCLKL tCLKH fCLK tCLKL tCLKH | 8.192 27 27 12.288 20 20 16.384 15 15 24.576 10 10 |  | 12.288 18.432 24.576 36.864 | MHz <br> ns <br> ns <br> MHz <br> ns <br> ns <br> MHz <br> ns <br> ns <br> MHz <br> ns <br> ns |
| LRCKA/B/C Timing (Slave Mode) |  |  |  |  |  |
| Normal Speed Mode <br> Double Speed Mode (LRCK A, LRCK C) <br> Quad Speed Mode (LRCK A, LRCK C) <br> Duty Cycle | fsn <br> fsd <br> fsq <br> Duty | $\begin{gathered} \hline 32 \\ 64 \\ 120 \\ 45 \end{gathered}$ |  | $\begin{gathered} \hline 48 \\ 96 \\ 192 \\ 55 \end{gathered}$ | $\begin{gathered} \hline \mathrm{kHz} \\ \mathrm{kHz} \\ \mathrm{kHz} \\ \% \end{gathered}$ |
| LRCKB Timing (Master Mode) |  |  |  |  |  |
| LRCKB frequency Duty Cycle | fs Duty | 32 | 50 | 48 | $\begin{gathered} \mathrm{kHz} \\ \% \end{gathered}$ |


| Parameter | Symbol | min | typ | max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Audio Interface Timing (Slave Mode) |  |  |  |  |  |
| PORTA, C |  |  |  |  |  |
| BICKA, C Period | tBCK | 81 |  |  | ns |
| BICKA,C Pulse Width Low | tBCKL | 32 |  |  | ns |
| Pulse Width High | tBCKH | 32 |  |  | ns |
| LRCKA,C Edge to BICKA " $\uparrow$ " (Note 24) | tLRB | 20 |  |  | ns |
| BICKA,C " $\uparrow$ " to LRCKA Edge (Note 24) | tBLR | 20 |  |  | ns |
| SDTIA, C Hold Time | tSDH | 10 |  |  | ns |
| SDTIA, C Setup Time | tSDS | 10 |  |  | ns |
| PORTB |  |  |  |  |  |
| BICKB Period | tBCK | 324 |  |  | ns |
| BICKB Pulse Width Low | tBCKL | 128 |  |  | ns |
| Pulse Width High | tBCKH | 128 |  |  | ns |
| LRCKB Edge to BICKB "个" (Note 24) | tLRB | 80 |  |  | ns |
| BICKB " $\uparrow$ " to LRCKB Edge (Note 24) | tBLR | 80 |  |  | ns |
| LRCKB to SDTOB1,2 (MSB) | tLRS |  |  | 80 | ns |
| BICKB " $\downarrow$ " to SDTOB1,2 | tBSD |  |  | 80 | ns |
| Audio Interface Timing (Master Mode) |  |  |  |  |  |
| BICKB Frequency | fBCK |  | 64fs |  | Hz |
| BICKB Duty | dBCK |  | 50 |  | \% |
| BICKB " $\downarrow$ " to LRCKB Edge | tMBLR | -40 |  | 40 | ns |
| BICKB " $\downarrow$ " to SDTO | tBSD |  |  | 20 | ns |
| Control Interface Timing ( $\mathbf{I}^{\mathbf{2}} \mathbf{C}$ Bus): |  |  |  |  |  |
| SCL Clock Frequency | fSCL | - |  | 400 | kHz |
| Bus Free Time Between Transmissions | tBUF | 1.3 |  | - | $\mu \mathrm{s}$ |
| Start Condition Hold Time (prior to first clock pulse) | thD:STA | 0.6 |  | - | $\mu \mathrm{s}$ |
| Clock Low Time | tLOW | 1.3 |  | - | $\mu \mathrm{s}$ |
| Clock High Time | tHIGH | 0.6 |  | - | $\mu \mathrm{s}$ |
| Setup Time for Repeated Start Condition | tSU:STA | 0.6 |  | - | $\mu \mathrm{s}$ |
| SDA Hold Time from SCL Falling (Note 25) | thD:DAT | 0 |  | - | $\mu \mathrm{s}$ |
| SDA Setup Time from SCL Rising | tSU:DAT | 0.1 |  | - | $\mu \mathrm{s}$ |
| Rise Time of Both SDA and SCL Lines | tR | - |  | 0.3 | $\mu \mathrm{s}$ |
| Fall Time of Both SDA and SCL Lines | tF | - |  | 0.3 | $\mu \mathrm{s}$ |
| Setup Time for Stop Condition | tSU:STO | 0.6 |  | - | $\mu \mathrm{s}$ |
| Pulse Width of Spike Noise Suppressed by Input Filter | tSP | - |  | 50 | ns |
| Capacitive load on bus | Cb | 0 |  | 400 | pF |
| Power-down \& Reset Timing |  |  |  |  |  |
| PDN Pulse Width (Note 27) | tPD | 150 |  |  | ns |
| PDN " $\uparrow$ " to SDTOB1,2 valid (Note 28) | tPDV |  | 522 |  | 1/fs |

Note 23. MCB supports only normal mode (256fsn, 384 fsn , 512 fsn , 768 fsn ).
Note 24. BICKA/B/C rising edge must not occur at the same time as LRCKA/B/C/ edge.
Note 25. Data must be held long enough to bridge the 300 ns -transition time of SCL.
Note 26. I ${ }^{2} \mathrm{C}$-bud is a trademark of NXP B.V.
Note 27. The AK4685 is reset by bringing the PND pin = "L".
Note 28. This is the number of LRCKB rising from PDN rising.

## ■ Timing Diagram



Clock Timing
LRCK $=$ LRCKA, LRCKB, LRCKC BICK= BICKA, BICKB, BICKC SDTI= SDTIA, SDTIC $\mathrm{SDTO}=\mathrm{SDTOB} 1, \mathrm{SDTOB} 2$.



## OPERATION OVERVIEW

## ■ System Clock

The AK4685 has three audio serial interfaces (PORTA, PORTB and PORTC) which can be operated asynchronously. The PORTA and PORTC are the audio data interfaces for DAC1 and DAC2, the PORTB is for ADC. At each PORT, the external clocks, which are required to operate the AK4685 in slave mode, are MCLKA, MCB, LRCKA/B/C, and BICKFA/B/C. The MCLKA/C and MCB must be synchronized with LRCKA/B/C but the phase is not critical.

## ■ Master/Slave Mode

The MSB pin selects master/slave mode of PORTB. PORTA and PORTC are slave mode only. In master mode, LRCKB pin and BICKB pin are output pins. In slave mode, LRCKA/B/C pins and BICKA/B/C pins are input pins (Table 1).

The PORTB is in slave mode at power-down (PDN pin = "L"). To change it to master mode, set the MSB pin to "H". Until setting the MSB pin to "H", LRCKB and BICKB pins are input pins. Around a 100kohm Pull-up (or down) resistor is required to prevent floating of these input pins.

| PDN pin | MSB pin | PORTB (ADC) <br> BICKB, LRCKB | PORTA/C (DAC1/2) <br> BICKA/C, LRCKA/C |
| :---: | :---: | :---: | :---: |
|  | L | Input (slave mode) | Input (slave mode) |
|  | H | Output "L"(master mode) | Input (slave mode) |
| H | L | Input (slave mode) | Input (slave mode) |
|  | H | Output (master mode) | Input (slave mode) |

Table 1. Master/Salve Mode

## AsahiKASEI

## ■ Cristal Oscillator Circuit

The clock for the MCB/XTI pin can be generated by the two methods:

1) $X^{\prime}$ tal


Note 29. The capacitor value is depend on the crystal oscillator (Typ.10-40pF)
Figure 2. X'tal Mode
2) External Clock

- Note: The clock must not over DVDD.


Figure 3. Direct Input

## ADC Clock Control

The integrated ADC of the AK 4685 operates by the clock from MCB/XTI pin.
In master mode (MSB pin = "H"), the CKS11-0 bits select the clock frequency (Table 2). The ADC is in power-down mode until MCB is supplied.

| CKSB1 | CKSB0 | Clock Speed |
| :---: | :---: | :---: |
| 0 | 0 | 256 fs |
| 0 | 1 | 384 fs |
| 1 | 0 | 512 fs |
| 1 | 1 | 768 fs |

Table 2. PORT1 Master Clock Control (ADC Master Mode)
In slave mode (MSB pin = "L"), the master clock (MCB) must be synchronized with LRCKB but the phase is not critical. After exiting reset state when power-up the device or other situations ( PDN pin $=$ " H "), the ADC is in power-down mode until MCB is input.

| LRCKB | MCB (MHz) |  |  |  |  |  | Sampling <br> Speed |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| fs | 128 fs | 192 fs | 256 fs | 384 fs | 512 fs | 768 fs |  |
| 32.0 kHz | - | - | 8.1920 | 12.2880 | 16.3840 | 24.5760 | Normal |
| 44.1 kHz | - | - | 11.2896 | 16.9344 | 22.5792 | 33.8688 |  |
| 48.0 kHz | - | - | 12.2880 | 18.4320 | 24.5760 | 36.8640 |  |

Table 3. System Clock Example (ADC Slave Mode)

## ■ DAC1/2 Clock Control

The master clock MCLKA (MCLKC) must be synchronized with LRCKA (LRCKC) but the phase is not critical. After exiting reset state when power-up the device or other situations ( PDN pin $=$ " H "), the DAC is in power-down mode until MCLKA/C and LRCKA/C are input.

There are two modes for controlling the sampling speed of DAC1(DAC2). One is the Manual Setting Mode (ACKS bit = " 0 ") using the DFS1-0 bits, and the other is Auto Setting Mode (ACKS bit $=$ " 1 ").

## 1. Manual Setting Mode (ACKS1(ACKS2) bit = "0")

When the ACKS1(ACKS2) bit $=$ " 0 ", DAC1(DAC2) is in Manual Setting Mode and the sampling speed is selected by DFS11-10, DFS21-20 bits (Table 4).

| DFS11 <br> (DFS21) | DFS10 <br> (DFS20) | DAC1(DAC2) Sampling Speed fs |  |
| :---: | :---: | :---: | :---: |
| 0 | 0 | Normal Speed Mode | $32 \mathrm{kHz} \sim 48 \mathrm{kHz}$ |
| 0 | 1 | Double Speed Mode | $64 \mathrm{kHz} \sim 96 \mathrm{kHz}$ |
| 1 | 0 | Quad Speed Mode | $120 \mathrm{kHz} \sim 192 \mathrm{kHz}$ |
| 1 | 1 | Not Available | - |

(Note: ADC is always in Normal Speed Mode)
Table 4. DAC Sampling Speed $($ ACKS $1 / 2$ bit $=" 0 "$, Manual Setting Mode $)$

| LRCKA/C | MCLKA/C (MHz) |  |  |  | BICKA/C (MHz) |
| :---: | :---: | :---: | :---: | :---: | :---: |
| fs | 256 fs | 384 fs | 512 fs | 768 fs | 64 fs |
| 32.0 kHz | 8.1920 | 12.2880 | 16.3840 | 24.5760 | 2.0480 |
| 44.1 kHz | 11.2896 | 16.9344 | 22.5792 | 33.8688 | 2.8224 |
| 48.0 kHz | 12.2880 | 18.4320 | 24.5760 | 36.8640 | 3.0720 |

Table 5. DAC System Clock Example (DAC Normal Speed Mode @Manual Setting Mode)

| LRCKA/C | MCLKA/C (MHz) |  |  |  | BICKA/C (MHz) |
| :---: | :---: | :---: | :---: | :---: | :---: |
| fs | 128 fs | 192 fs | 256 fs | 384 fs | 64 fs |
| 88.2 kHz | 11.2896 | 16.9344 | 22.5792 | 33.8688 | 5.6448 |
| 96.0 kHz | 12.2880 | 18.4320 | 24.5760 | 36.8640 | 6.1440 |

Table 6. DAC System Clock Example (DAC Double Speed Mode @Manual Setting Mode)

| LRCKA/C | MCLKA/C (MHz) |  |  |  | BICKA/C (MHz) |
| :---: | :---: | :---: | :---: | :---: | :---: |
| fs | 128 fs | 192 fs | 256 fs | 384 fs | 64 fs |
| 176.4 kHz | 22.5792 | 33.8688 | - | - | 11.2896 |
| 192.0 kHz | 24.5760 | 36.8640 | - | - | 12.2880 |

Table 7. DAC system clock example (DAC Quad Speed Mode @Manual Setting Mode)

## 2. Auto Setting Mode (ACKS1/2 bit = " 1 ")

When the ACKS1(ACKS2) bit = " 1 ", the DAC is in Auto Setting Mode and the sampling speed is selected automatically by the ratio of MCLKA/LRCKA or MCLKC/LRCKC, as shown in the Table 8 and Table 9. In this mode, the settings of DFS21-20 bit or FS11-10 bit are ignored.

| MCLKA/C | DAC1/2 Sampling Speed (fs) LRCKA/C |  |
| :---: | :---: | :---: |
| $512 \mathrm{fs}, 768 \mathrm{fs}$ | Normal Speed Mode | $32 \mathrm{kHz} \sim 48 \mathrm{kHz}$ |
| $256 \mathrm{fs}, 384 \mathrm{fs}$ | Double Speed Mode | $64 \mathrm{kHz} \sim 96 \mathrm{kHz}$ |
| $128 \mathrm{fs}, 192 \mathrm{fs}$ | Quad Speed Mode | $120 \mathrm{kHz} \sim 192 \mathrm{kHz}$ |

(Note: ADC is always in Normal Speed Mode)
Table 8. DAC Sampling Speed (ACKS bit $=" 1 "$, Auto Setting Mode)

| LRCKA/C | MCLKA/C (MHz) |  |  |  |  |  | $\begin{array}{c}\text { Sampling } \\ \text { Speed }\end{array}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $y y y y y y y y n n$ | fs | 128 fs | 192 fs | 256 fs | 384 fs | 512 fs |  |$]$

Table 9. DAC System Clock Example (Auto Setting Mode)

## ■ De-emphasis Filter

The AK4685 includes a digital de-emphasis filter ( $\mathrm{tc}=50 / 15 \mu \mathrm{~s}$ ) by IIR filter. This filter corresponds to three sampling frequencies ( $32 \mathrm{kHz}, 44.1 \mathrm{kHz}, 48 \mathrm{kHz}$ ). De-emphasis filter is off in Double speed mode and Quad speed mode. Deemphasis of each DAC can be set individually by register.

| Mode | Sampling Speed | DEM21/DEM11 | DEM20/DEM10 | DEM |
| :---: | :---: | :---: | :---: | :---: |
| 0 | Normal Speed | 0 | 0 | 44.1 kHz |
| 1 | Normal Speed | 0 | 1 | OFF |
| 2 | Normal Speed | 1 | 0 | 48 kHz |
| 3 | Normal Speed | 1 | 1 | 32 kHz |

Table 10. De-emphasis control

## ■ ADC Digital High Pass Filter

The integrated ADC has a digital high pass filter for DC offset cancelling. The cut-off frequency is 1.0 Hz at $\mathrm{fs}=48 \mathrm{kHz}$ and scales with sampling rate (fs).

## Audio Serial Interface Format

Each PORTA/B/C can select independent audio interface format. DIFA1-0 bits control the PORTA. The MSB pin and DIFB bit control PORTB. In all modes, the serial data is MSB first, 2's complement format. The SDTOB1/2 pins are clocked out on the falling edge of BICKB pin and the SDTIA/C pins are latched on the rising edge of BICKA/C pins. " 0 " should be written to LSB bits without data on each SDTIA/C input.

## 1. PORTA/C Setting

The DIFA1-0 bits and DIFC1-0 bits select following four serial data formats (Table 11).

| Mode | DIFA1 <br> (DIFC1) <br> bit | DIFA0 <br> (DIFC0) <br> bit |  | SDTIA1 |  | LRCKA |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | BICKA |  |  |  |  |
| 0 | 0 | 0 | 20bit, Right justified | H/L | I | $\geq 48 \mathrm{fs}$ | I |
| 1 | 0 | 1 | 24bit, Right justified | H/L | I | $\geq 48 \mathrm{fs}$ | I |
| 2 | 1 | 0 | 24bit, Left justified | H/L | I | $\geq 48 \mathrm{fs}$ | I |
| 3 | 1 | 1 | 24bit, I $\mathrm{I}^{2} S$ | L/H | I | $\geq 48 \mathrm{fs}$ | I |

(default)
Table 11. Audio Interface Format

## 2. PORTB Setting

2-1: Normal mode:
MSB pin and DIFB bit select following four serial data formats (Table 12).

| Mode | $\begin{gathered} \text { MSB } \\ \text { pin } \end{gathered}$ | $\begin{gathered} \hline \text { DIFB } \\ \text { bit } \\ \hline \end{gathered}$ | SDTOB1,2 | LRCKA |  | BICKA |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | L/R | I/O | speed | I/O |
| 0 | L | 0 | 24bit, Left justified | H/L | I | $\geq 48 \mathrm{fs}$ | I |
| 1 | L | 1 | 24bit, $\mathrm{I}^{2} \mathrm{~S}$ | L/H | I | $\geq 48 \mathrm{fs}$ | I |
| 2 | H | 0 | 24bit, Left justified | H/L | O | 64fs | O |
| 3 | H | 1 | 24bit, I ${ }^{2}$ S | L/H | O | 64fs | O |

Table 12. Audio Interface Format


Figure 5. Audio Data Timing (SDTO: Mode0/2, SDTI: Mode0)


Figure 6. Audio Data Timing (SDTO: Mode0/2, SDTI: Mode1)


Figure 7. Audio Data Timing (SDTO: Mode0/2, SDTI: Mode2)


Figure 8. Audio Data Timing (SDTO: Mode1/3, SDTI: Mode3)

## Digital Volume Control

The AK4685 has channel-independent digital volume control ( 256 levels, 0.5 dB step). The IATL7-0, IATR7-0 bits set the volume level of ADC channel (Table 13). The OAT1L7-0, OAT1R7-0, OAT2L7-0 and OAT2R7-0 bits set each DAC channel (Table 14).

| IATL7-0, <br> IATR7-0 | Gain |
| :---: | :---: |
| 00 H | +24 dB |
| 01 H | +23.5 dB |
| 02 H | +23.0 dB |
| $:$ | $:$ |
| 2 FH | +0.5 dB |
| 30 H | 0 dB |
| 31 H | -0.5 dB |
|  | $:$ |
| FEH | -103 dB |
| FFH | MUTE $(-\infty)$ |

(default)

Table 13. ADC Digital Volume (IATT)

| OAT1L7-0, <br> OAT1R7-0, <br> OAT2L7-0, <br> OAT2R7-0 | Gain |
| :---: | :---: |
| 00 H | +12 dB |
| 01 H | +11.5 dB |
| 02 H | +11.0 dB |
| $:$ | $:$ |
| 17 H | +0.5 dB |
| 18 H | 0 dB |
| 19 H | -0.5 dB |
|  | $:$ |
| FEH | -115 dB |
| FFH | MUTE $(-\infty)$ |

Table 14. DAC Digital Volume (OATT)
ATSAD (ATSDA) bits (Table 15, Table 16) control the transition time of attenuation. The transition between each attenuation level is the soft transition. Therefore, the switching noise does not occur in the transition.

| Mode | ATSAD | ATT speed |
| :---: | :---: | :---: |
| 0 | 0 | $1061 / \mathrm{fs}$ |
| 1 | 1 | $256 / \mathrm{fs}$ |

Table 15. Transition time of attenuation (ADC)

| Mode | ATSDA | ATT speed |
| :---: | :---: | :---: |
| 0 | 0 | $1061 / \mathrm{fs}$ |
| 1 | 1 | $256 / \mathrm{fs}$ |

Table 16. Transition time of attenuation (DAC1/2)

The transition between set values is soft transition of 1061 levels in Mode 0 . It takes $1061 / \mathrm{fs}(22 \mathrm{~ms} @ \mathrm{fs}=48 \mathrm{kHz})$ from 00 H to $\mathrm{FFH}(\mathrm{MUTE})$. If the PDN pin goes to "L", the IATL7-0, IATR7-0 (OAT1L7-0, OAT1R7-0, OAT2L7-0, OAT2R7-0) bits are initialized to $30 \mathrm{H}(18 \mathrm{H})$. The ATT levels go to their default value when RSTN bit $=$ " 0 ". When RSTN bit return to " 1 ", the ATTs fade to their current value.

## ■ Digital Soft Mute

The ADC and DAC have a soft mute function. The soft mute operation is performed at digital domain. When the SMAD/SMDA bits go to " 1 ", the output signal is attenuated by $-\infty$ during ATT_DATA $\times$ ATT transition time (Table 15, Table 16) from the current ATT level. When the SMAD/SMDA bits are returned to " 0 ", the mute is cancelled and the output attenuation gradually changes to the ATT level during ATT_DATA×ATT transition time. If the soft mute is cancelled before attenuating to $-\infty$ after starting the operation, the attenuation is discontinued and returned to ATT level in the same cycle. The soft mute is effective for changing the signal source without stopping the signal transmission.


Notes:
(1) ATT_DATA×ATT transition time (Table 15, Table 16). For example, in Normal Speed Mode, this time is $1061 / \mathrm{fs}$ cycles $(256 / \mathrm{fs})$ at ATT_DATA $=00 \mathrm{H}$. ATT transition of the soft-mute is from 00 H to FFH
(2) The analog output corresponding to the digital input has group delay, GD.
(3) If the soft mute is cancelled before attenuating to $-\infty$ after starting the operation, the attenuation is discontinued and returned to ATT level in the same cycle.

Figure 9. Soft Mute Function

## ■ Pre-Amp and Differential Inputs

The input ATTs are constructed by connecting input resistors (Ri) to LIN1+/- pins and feedback resistors (Rf) between LAMPO1/ LVCOM1 pin and LIN1-/LIN1+ pin (Figure 10). The input voltage range of the LAMPO1/ LVCOM1 pin is typically $+/-0.33 \times$ AVDD (Vpp). If the input voltage of the input selector exceeds typ. $+/-5.66 \mathrm{Vpp}(+/-2 \mathrm{Vrms}$ differential), or $+/-8.48 \mathrm{Vpp}(+/-3 \mathrm{Vrms}$ differential), the input voltage of the LAMPO1/ LVCOM1 pins must be attenuated to typ. +/-3.3 x AVDD1/5 (Vpp) ATTs. Table 17 shows Ri and Rf constant examples.


Figure 10. External Connection Example (differential input)

| Input Range | $\operatorname{Ri}[\mathrm{k} \Omega]$ | $\operatorname{Rf}[\mathrm{k} \Omega]$ | ATT Gain [dB] | Voltage between LAMPO1 <br> and LVCOM1 pins |
| :---: | :---: | :---: | :---: | :---: |
| $+/-8.48 \mathrm{Vpp}$ <br> $(+/-3 \mathrm{Vrms} \mathrm{Differential} \mathrm{Input)}$ | 47 | 18 | -8.3 | $+/-3.25 \mathrm{Vpp}$ |
| +/-5.66Vpp <br> $(+/-2 \mathrm{Vrms}$ Differential Input $)$ | 33 | 18 | -5.3 | $+/-3.08 \mathrm{Vpp}$ |
| $+/-2.83 \mathrm{Vpp}$ <br> $(+/-1 \mathrm{Vrms}$ Differential Input) | 16 | 18 | +1.02 | $+/-3.18 \mathrm{Vpp}$ |

Note 30. The input range of the internal ADC is +/-3.3 x AVDD1/5 Vpp typ.
Note 31. The input range is the voltage difference of Ri inputs (AINL1/L2/R1/R2+)-(AINL1/L2/R1/R2-).
Table 17. Input ATT Example (differential input)
(Pseudo Cap-less /Single-ended input)
The input ATTs are constructed by connecting input resistors (Ri) to LIN1+/- pins and feedback resistors (Rf) between LAMPO1/ LVCOM1 pin and LIN1-/LIN1+ pin (Figure 11) when using single-ended and pseudo cap-less inputs as well as when using differential inputs. The input voltage range of the LAMPO1/ LVCOM1 pin is typically $+/-0.33 \mathrm{x}$ AVDD (Vpp). If the input voltage of the input selector exceeds typ. $5.66 \mathrm{Vpp}(2 \mathrm{Vrms})$ or $8.48 \mathrm{Vpp}(3 \mathrm{Vrms})$, the input voltage of the LAMPO1/ LVCOM1 pins must be attenuated to typ. +/-3.3 x AVDD1/5 (Vpp) ATTs. Table 18 shows Ri and Rf constant examples.


Figure 11. External Connection Example (single-ended input)

| Input Range | $\operatorname{Ri}[\mathrm{k} \Omega]$ | $\operatorname{Rf}[\mathrm{k} \Omega]$ | ATT Gain [dB] | Voltage between LAMPO1 <br> and LVCOM1 pins |
| :---: | :---: | :---: | :---: | :---: |
| 8.48 Vpp <br> $(3 \mathrm{Vrms})$ | 47 | 36 | -2.3 | $+/-3.25 \mathrm{Vpp}$ |
| 5.66 Vpp <br> $(2 \mathrm{Vrms})$ | 33 | 36 | +0.7 | $+/-3.08 \mathrm{Vpp}$ |
| 2.83 Vpp <br> $(1 \mathrm{Vrms})$ | 16 | 36 | +7.02 | $+/-3.18 \mathrm{Vpp}$ |

Note 32. The input range of the internal ADC is +/-3.3 x AVDD1/5 Vpp typ.
Note 33. The input range is the voltage difference of Ri inputs (AINL1/L2/R1/R2)-GND.

Table 18. Input ATT example (single-end input)

## Analog Outputs

(Differential Mode)
The analog outputs are fully differential outputs when the SGL pin = "L", and the output range is $2.56 \times($ AVDD2 $) / 5$ Vpp centered around analog common voltage (VCOM pin). The differential outputs are summed externally. The summing gain between $\mathrm{L} /$ ROUT + and $\mathrm{L} /$ ROUT- is $\mathrm{V}_{\mathrm{L} / \text { ROUT }}=(\mathrm{L} /$ ROUT +$)-(\mathrm{L} /$ ROUT -$)$. If the summing gain is 1.09 , the output range is $5.59 \mathrm{Vpp}(\operatorname{typ} @ \mathrm{VDD}=5 \mathrm{~V})$. The bias voltage of the external summing circuit is supplied externally. The output voltage ( $\mathrm{V}_{\text {AOUT }}$ ) is positive full scale for 7FFFFFH (@24-bits) and negative full scale for 800000 H (@24-bits). The ideal $\mathrm{V}_{\text {AOUT }}$ is 0 V for $000000 \mathrm{H}(@ 24$-bits).

The internal switched-capacitor filters attenuate the noise generated by the delta-sigma modulator beyond the audio passband. Figure 12 and Figure 13 show examples of an external LPF circuit summing the differential outputs with an op-amp.


Figure 12. External $2^{\text {nd }}$ order LPF Circuit Example (using op-amp with single power supply)


Figure 13. External $2^{\text {nd }}$ order LPF Circuit Example (using op-amp with dual power supplies)

## (Single-ended Mode)

The analog outputs are single-ended when the SGL pin = "H" and the signals are output from the L/ROUT + pins. In this case, the L/ROUT- pins should be opened. The output range is $2.8 \times$ (AVDD2)/5 Vpp (typ) centered around the analog common voltage (VCOM pin).

## - Charge Pump Circuit

The internal charge pump circuit generates negative voltage (PVEE) from PVDD voltage for headphone amplifiers. The internal charge pump starts operation when PWDA2 bit = " 1 ".

The power up time of charge pump circuit is maximum 8.0 ms . When PWHP bits $=$ " 1 ", the Headphone-Amp is powered-up after the charge pump circuit is powered-up.

## ■ Headphone-Amp (HPL/HPR pins)

Power supply voltage for headphone amplifiers is applied from a regulator for positive power and a charge-pump for negative power. The Regulator is driven by AVDD3 and the charge-pump is driven by PVDD. The PVEE pin outputs the negative voltage generated by the internal charge pump circuit. The headphone amplifier output is single-ended and centered on 0 V (VSS5). Therefore, the capacitor for AC-coupling can be removed. The minimum load resistance is $32 \Omega$. When the DAC input signal level is 0 dBFS , the output voltage is $1.21 \mathrm{Vrms}(=31 \mathrm{~mW} @ 32 \Omega$ via 6.8 ohm resistor) at HPGA4-0 bits $=0 \mathrm{~dB}$. The output level of headphone-amp can be controlled by HPGA4-0 bits. This volume setting is common to $\mathrm{L} / \mathrm{R}$ channels and can attenuate / gain the mixer output from +12 dB to -50 dB in 2 dB step. When changing the volume, pop noise occurs.

| HPGA4-0 bits | GAIN (dB) | Step |
| :---: | :---: | :---: |
| 1 FH | +12 |  |
| 1 EH | +10 |  |
| $:$ | $:$ |  |
| 1 AH | +2 |  |
| 19 H | 0 |  |
| 18 H | -2 |  |
| 17 H | -4 |  |
| 16 H | -6 |  |
| $:$ | $:$ |  |
| 2 H | -46 |  |
| 1 H | -48 |  |
| 0 H | -50 |  |
| (default) |  |  |

Table 19. Headphone-Amp Volume Setting

When PWHP bit is " 1 ", the headphone-amps are powered-up. The headphone output is enabled when HPMTN bit is " 1 " and muted when HPMTN bit is " 0 ". The mute ON/OFF time is set by PTS1-0 bits when MOFF bit is " 0 ".

| PTS2 | PTS1 | PTS0 | MUTE ON/OFF Time |  |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | typ. | max. |
| 0 | 0 | 0 | (reserved) | (reserved) |
| 0 | 0 | 1 | (reserved) | (reserved) |
| 0 | 1 | 0 | 4.1 ms | 6.9 ms |
| 0 | 1 | 1 | 8.2 ms | 13.9 ms |
| 1 | 0 | 0 | 16.4 ms | 27.7 ms |
| 1 | 0 | 1 | 32.8 ms | 55.4 ms |
| 1 | 1 | 0 | 65.6 ms | 100.8 ms |
| 1 | 1 | 1 | 131.2 ms | 221.6 ms |

(default)

Table 20. Headphone-Amp Mute ON/OFF Transition Time

Soft transition Enable/Disable is controlled by MOFF bit. When this bit is " 1 ", soft transition is disabled and the headphone is switched ON/OFF immediately. When soft transition is enabled, a register setting of the address 0BH should be made in an interval more than soft transition time. Register writings are ignored if the same value is written to these registers.

When PWHP bit is " 0 ", the headphone-amps are powered-down completely. At that time, the HPL and HPR pins go to VSS5 voltage via the internal pulled-down resistor. The pulled-down resistor is $20 \Omega(\mathrm{typ})$ at HPZ bit $=$ " 0 ", $50 \mathrm{k} \Omega(\mathrm{typ})$ at HPZ bit $=$ " 1 ". The power-up time is 16.4 ms (typ.) and 27.7 ms (max.), and power up/down is executed immediately.

| PWHP | HPZ | PWDA2 | HPMTN | Mode | HPL/R pin states |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | x | x | Power-down \& Mute | Pulled-down by $20 \Omega$ (typ) |
| 0 | 1 | 0 | x | N/A | (default) |
| 0 | 1 | 1 | x | Power-down | Pull-down by $50 \mathrm{k} \Omega$ (typ) |
| 1 | x | 0 | x | N/A | N/A |
| 1 | x | 1 | 0 | Mute | VSS5 |
| 1 | x | 1 | 1 | Normal Operation | Normal Operation |

Table 21. Headphone Outputs Status (x: Don't' care)

## ■ Clock Stop Detection Function

When MCLKA, MCB and MCLKC external clocks are stopped, corresponding digital blocks become power-down mode.
The power-down mode is released automatically and digital blocks return to normal operation when external clocks are supplied again. An initialization cycle of $522 / \mathrm{fs}$ is taken before returning to normal operation when MCB clock is stopped.

## Analog Mute

When the MT1N pin is set to "L" from "H", a digital to analog data converting is stropped and the analog outputs (LOUT+/-, ROUT+/- pins) are attenuated in soft transition. The analog block becomes power-down mode after the soft transition is completed, and VCOM is output from the analog outputs. Transition time is controlled by AMT2-0 bits. When the MT1N pin is set to "H" from "L", the analog block returns to normal operation and a digital to analog converting is resumed. After DAC initializing time, the mute is cancelled and the output attenuation gradually changes to the ATT level during ATT_DATA x ATT transition time. (Table 16)

(1) "L" time of 20 ms or more is needed.
(2) The soft mute transition time by analog processing is depending on the AMT2-0 bits setting. A crick noise occurs when each power supply (TVDD, DVDD $1 / 2 / 3$, AVDD $1 / 2 / 3$ and PVDD) is off during mute transition time. Power supplies should be provided longer than the transition time set by AMTS2-0 bits set.
(3) ATT_DATA x ATT transition time (Table 16). In case of MODE0 and ATS2-0 bits $=$ " $00 \mathrm{H} "$, the transition time of ATT value from $\mathrm{FFH}(0 \mathrm{~dB})$ to 00 H (MUTE) is $1061 / \mathrm{fs}$.

Figure 14. Mute Sequence Example (MT1N pin)

When the MT2N pin is set to "L" from "H", the headphone output is attenuated in soft transition. The analog and headphone blocks become power-down mode after the soft transition is completed, and ground level (VSS5) is output from these outputs. Transition time is controlled by AMT2-0 bits. The data inputs and DAC clocks must not be stopped before the soft transition complete.
When the MT2N pin is set to "H" from "L", the analog and headphone blocks return to normal operation and a digital to analog converting is resumed. After DAC initializing time, the mute is cancelled and the output attenuation gradually changes to the ATT level during ATT_DATA x ATT transition time. (Table 16)

(1) "L" time of 20 ms or more is needed.
(2) The mute time of the headphone amplifier is $8.2 \mathrm{~ms}(\mathrm{typ})$ and $14 \mathrm{~ms}(\max )$ when PTSA bit $=$ " 0 " (at default PTS2-0 bits $=$ " 011 "). PST2-0 bits setting does not effect this mute time. The mute time can be controlled by PTS2-0 bits setting when PTSA bit $=$ " 1 ".
A crick noise occurs when each power supply (TVDD, DVDD1/2/3, AVDD1/2/3 and PVDD) is off during mute transition time. Power supplies should be provided longer than the transition time set by AMTS2-0 bits set.
(3) ATT_DATA x ATT transition time (Table 16). In case of MODE0 and ATS2-0 bits $=$ " $00 \mathrm{H} "$ ", the transition time of ATT value from $\mathrm{FFH}(0 \mathrm{~dB})$ to 00 H (MUTE) is $1061 / \mathrm{fs}$.
(4) Power down time of the headphone amplifier is controlled by AMTS2-0 bits. The AMTS2-0 bits setting value should be shorter than PTS2-0 bits setting value.
(5) Headphone amplifier power-up: GND level is output when the headphone amplifier is muted. The headphone amplifier power-up time is 27.7 ms (max).
(6) The mute release time of the headphone amplifier is controlled by PTS2-0 bits or MOFF bit settings.

Figure 15. Mute Sequence Example (MT2N pin)

AMTS2-0: Analog MUTE Power-down time control

| AMTS | AMTS | AMTS | Power-down time |  |
| :---: | :---: | :---: | :---: | :---: |
| 2 | 1 | 0 | typ. | max. |
| 0 | 0 | 0 | 10 ms | 17 ms |
| 0 | 0 | 1 | 21 ms | 35 ms |
| 0 | 1 | 0 | 41 ms | 70 ms |
| 0 | 1 | 1 | 82 ms | 140 ms |
| 1 | 0 | 0 | 164 ms | 280 ms |
| 1 | 0 | 1 | 5.1 ms | 8.6 ms |
| 1 | 1 | $X$ | 1.3 ms | 2.2 ms |

Table 22. Power-down Time Control

L/R channels of the analog outputs (LOUT $+/$-, ROUT $+/-$ pins) can be muted independently by AMT1LN or AMT1RN bits $=$ " 0 ". When those channels are muted, transition time is depending on AMT2-0 bits setting. Each mute is cancelled by AMT1LN / AMT1RN bit = " 1 ". It is digitally-processed, and the output attenuation gradually changes to the ATT level during ATT_DATA x ATT transition time. (Table 16)

(1) "L" time of 20 ms or more is needed.
(2) The soft mute transition time by analog processing is depending on the AMTS2-0 bits setting.
(3) ATT_DATA x ATT transition time (Table 16). In case of MODE0 and ATS2-0 bits =" $00 \mathrm{H} "$, the transition time of ATT value from $\mathrm{FFH}(0 \mathrm{~dB})$ to 00 H (MUTE) is $1061 / \mathrm{fs}$.

Figure 16. Mute Sequence Example (AMT1LN/1RN bit)

When AMT1LN=AMT1RN bit = " 0 ", the analog outputs (DACL+/-, DACR+/- pins) are attenuated in soft transition as well as when the MT1N pin = "L". The analog block becomes power-down mode after the soft transition is completed, and VCON is output from the analog outputs.
When a one of or both AMT1L and AMT1RN bits are set to " 1 ", the analog block returns to normal operation. The mute on the corresponding channel is cancelled and the output attenuation gradually changes to the ATT level during ATT_DATA x ATT transition time. (Table 16)

When the AMT2LN=AMT2RN bit = " 0 ", the analog and headphone blocks become power-down mode, and ground level (VSS5) is output from these outputs as well as when the MT2N pin = "L". Transition time is controlled by AMT2-0 bits. The data inputs and DAC clocks must not be stopped before the soft transition complete.
When $t$ both AMT2L and AMT2RN bits are set to " 1 ", the analog and headphone blocks return to normal operation. The mute is cancelled and the output attenuation gradually changes to the ATT level during ATT_DATA x ATT transition time. (Table 16)

## ■ Power ON/OFF Sequence

The each block of the AK4685 is placed in power-down mode by bringing the PDN pin "L" and both digital filters are reset at the same time. The PDN pin ="L" also reset the control registers to their default values. In power-down mode, the DAC1/2 outputs go to VSS3/5 and the SDTOB1/2 pins go to "L". The AK4685 should be powered-up when the PDN pin = "L" to reset the internl registers.
In slave mode, after exiting reset at power-up or other situations, the $\mathrm{ADC} / \mathrm{DAC} 1 / \mathrm{DAC} 2$ starts operation on the rising edge of LRCKB/A/C after MCB/MLCKA/C inputs. The ADC is in power-down mode until MCB is input, and the $\mathrm{DAC} 1 / 2$ are in power-down mode until MLCKA/C or LRCKA/C is input.
The analog initialization cycle of ADC starts after exiting the power-down mode. Therefore, the output data, SDTOB1/2 becomes available after $522 /$ fs cycles of LRCKB clock. In case of the $\mathrm{DAC1} / 2$, an analog initialization cycle starts after exiting the power-down mode. The analog outputs are VSS3/5 during the initialization. Figure 17 shows the sequences of the power-down and the power-up.
The ADC and DAC's can be powered-down individually by PWAD and PWDA1/2 bits. These bits do not initialize the internal register values. When PWAD bit $=$ " 0 ", the SDTOB1/2 pins go to "L". When PWDA1 bit $=$ " 0 ", the analog outputs (LOUT $+/-$, ROUT $+/-$ pins) go to VCOM voltage. When PWDA2 bit $=$ " 0 ", the headphone outptus (HPL/R pins) go to VSS5 voltage. As some click noise occurs, the analog output should be muted externally if the click noise influences a system application.


Figure 17. Power-up/down Sequence Example

## Notes:

(1) The PDN pin should be set "L" $\rightarrow$ "H" after the all powers (TVDD, DVDD1/2/3, AVDD1/2/3 and PVDD) are supplied. The AK4685 requires 150 ns or longer "L" period for a reset. The AK4685 should be powered-up when the PDN pin = "L".
(2) Power-on the regulator, charge pump circuit, VCOM, HP-Amp and internal oscillator

The PVEE pin becomes to the same voltage as PVEE within 8.0 ms (max).
(3) The analog block of the ADC is initialized after exiting the power-down state.
(4) The analog block of the DAC is initialized after exiting the power-down state.
(5) The digital outputs corresponding to analog inputs, and the analog outputs corresponding to digital inputs have group delay (GD).
(6) ADC output is " 0 " data at the power-down state.
(7) Click noise occurs at the end of initialization of the analog block. Mute the digital outputs externally if the click noise influences a system application.
(8) A click noise occurs at the falling edge of PDN and at $512 / \mathrm{fs}$ after the rising edge(after charge-pump is power-on) of PDN.
(9) Power-up of Headphone-Amp: PWHP bit $=$ " 0 " $\rightarrow$ " 1 "

Headphone-Amp is in mute state and outputs ground level. Headphone-Amp power-up time is 27.7 ms (max.).
(10) Headphone-Amp mute release: HPMTN pin = "L" $\rightarrow$ "H"

Headphone-Amp goes to the normal operation after the transition time. Headphone-Amp mute release time depends on the setting of PTS1-0 and MOFF bits.
(11) Headphone-Amp mute: HPMTN pin $=$ "H" $\rightarrow$ "L"

Headphone-Amp goes to mute state after the transition time set by PTS1-0 and MOFF bits.
(12) Headphone-Amp power-down: PWHPL/R bits $=$ " 1 " $\rightarrow$ " 0 "

Headphone-Amp is powered-down immediately.
(13) PWDA2 bit = " 1 " $\rightarrow$ " 0 "

The PVEE pin becomes 0 V according to the time constant of the capacitor at the PVEE pin and the internal resistor. The internal resistor is $17.5 \mathrm{k} \Omega$ (typ.).
(14) Mute the analog outputs externally if the click noise (8) influences a system application.

## Reset Function

When RSTN bit $=$ " 0 ", the ADC and DAC digital blocks are powered-down but the internal register are not initialized. The analog outputs (LOUT $+/-$, ROUT $+/-$ pins) go to VCOM voltage, the headphone outputs (HPL/R pins) go to ground level (VSS5) and the SDTOB1/2 pins go to "L". As some click noise occur, the analog outputs should be muted externally if the click noise influences a system application. The Figure 18 shows the power-up sequence.


Notes:
(1) The analog block of ADC is initialized after exiting the reset state.
(2) The digital outputs corresponding to the analog inputs, and the analog outputs corresponding to the digital inputs have group delay (GD).
(3) ADC output is " 0 " data at power-down state.
(4) Click noise occurs when the internal RSTN bit becomes " 1 ". Mute the digital outputs externally if the click noise influences a system application.
(5) When RSTN bit $=$ " 0 ", the analog outputs go to 0 V .
(6) A click noise occurs at $4 \sim 5 /$ fs after RSTN bit became " 0 ", and occurs at $1 \sim 2 / f s$ after RSTN bit becomes " 1 ".
(7) There is a delay about $4 \sim 5 / \mathrm{fs}$ from a writing " 0 " to the RSTN bit until the internal RSTN bit changes to " 0 ".

Figure 18. Reset Sequence Example

## Serial Control Interface

The AK4685 supports fast-mode $\mathrm{I}^{2} \mathrm{C}$-bus system (max: 400 kHz ).

## 1. Data transfer

All commands are preceded by START condition. After the START condition, a slave address is sent. After the AK4685 recognizes START condition, the device interfaced to the bus waits for the slave address to be transmitted over the SDA line. If the transmitted slave address matches an address for one of the devices, the designated slave device pulls the SDA line to LOW (ACKNOWLEDGE). The data transfer is always terminated by STOP condition generated by the master device.

## 1-1. Data validity

The data on the SDA line must be stable during the HIGH period of the clock. The HIGH or LOW state of the data line can only change when the clock signal on the SCL line is LOW except for the START and the STOP condition.


Figure 19. Data Transfer

## 1-2. START and STOP condition

A HIGH to LOW transition on the SDA line while SCL is HIGH indicates START condition. All sequences start from START condition. A LOW to HIGH transition on the SDA line while SCL is HIGH defines STOP condition. All sequences end by STOP condition.


Figure 20. START and STOP conditions

## 1-3. ACKNOWLEDGE

ACKNOWLEDGE is a software convention used to indicate successful data transfers. The transmitting device will release the SDA line $(\mathrm{HIGH})$ after transmitting eight bits. The receiver must pull down the SDA line during the acknowledge clock pulse so that SDA remains stable "L" during "H" period of this clock pulse. The AK4685 will generates an acknowledge after each byte has been received.
In read operation, the slave, the AK4685 will transmit eight bits of data, release the SDA line and monitor the line for an acknowledge. If an acknowledge is detected and no STOP condition is generated by the master, the slave will continue to transmit data. If an acknowledge is not detected, the slave will terminate further data transmissions and await STOP condition.


Figure 21. Acknowledge on the $\mathrm{I}^{2} \mathrm{C}$-bus

## 1-4. FIRST BYTE

The first byte, which includes seven bits of slave address and one bit of R/W bit, is sent after START condition. If the transmitted slave address matches an address for one of the device, the receiver who has been addressed pulls down the SDA line.
The most significant seven bits of the slave address are fixed as " 0010010 ". The eighth bit (LSB) of the first byte (R/W bit) defines whether a write or read condition which the master requests. " 1 " indicates that the read operation is to be executed. " 0 " indicates that the write operation is to be executed.

| 0 | 0 | 1 | 0 | 0 | 1 | 0 | $\mathrm{R} / \mathrm{W}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Figure 22. The First Byte

## 2. WRITE Operations

Set R/W bit = "0" for the WRITE operation of the AK4685.
After receipt of the start condition and the first byte, the AK4685 generates an acknowledge, and awaits the second byte (register address). The second byte consists of the address for control registers of AK4685. The format is MSB first, and those most significant 3-bits are "Don't care".

(*: Don’t care)
Figure 23. The Second Byte
After receipt of the second byte, the AK4685 generates an acknowledge, and awaits the third byte. Those data after the second byte contain control data. The format is MSB first, 8 bits.

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Figure 24. Byte structure after the second byte
The AK4685 is capable of more than one byte write operation in one sequence.
After a receipt of the third byte, the AK4685 generates an acknowledge, and awaits the next data again. The master can transmit more than one word instead of terminating the write cycle after the first data word is transferred. After the receipt of each data, the internal 5bits address counter is incremented by one, and the next data is taken into next address automatically. If the address exceeds 0 CH prior to generating the stop condition, the address counter will "roll over" to 00 H and the previous data will be overwritten.


Figure 25. WRITE Operation

## 3. READ Operations

Set R/W bit = " 1 " for the READ operation of the AK4685.
The master can read next address's data by generating the acknowledge instead of terminating the write cycle after the receipt of the first data word. After the receipt of each data, the internal 5 bits address counter is incremented by one, and the next data is taken into next address automatically. If the address exceeds 0 CH prior to generating stop condition, the address counter will "roll over" to 00 H and the previous data will be overwritten.

The AK4685 supports two basic read operations: CURRENT ADDRESS READ and RANDOM READ.

## 3-1. CURRENT ADDRESS READ

The AK4685 contains an internal address counter that maintains the address of the last word accessed, incremented by one. Therefore, if the last access (either a read or write) was to address " n ", the next CURRENT READ operation would access data from the address " $n+1$ ".
After receipt of the slave address with R/W bit set to " 1 ", the AK 4685 generates an acknowledge, transmits lbyte data which address is set by the internal address counter and increments the internal address counter by 1 . If the master does not generate an acknowledge to the data but generate stop condition, the AK4685 discontinues transmission


Figure 26. CURRENT ADDRESS READ

## 3-2. RANDOM READ

Random read operation allows the master to access any memory location at random. Prior to issuing the slave address with the R/W bit set to " 1 ", the master must first perform a "dummy" write operation.
The master issues start condition, slave address( $\mathrm{R} / \mathrm{W}$ bit=" 0 ") and then the register address to read. After the register address's acknowledge, the master immediately reissues the start condition and the slave address with the $\mathrm{R} / \mathrm{W}$ bit set to " 1 ". Then the AK 4685 generates an acknowledge, 1byte data and increments the internal address counter by 1 . If the master does not generate an acknowledge but generate the stop condition, the AK4685 discontinues transmission.


Figure 27. RANDOM READ

## ■ Register Map

| Addr | Register Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 00H | Powerdown | PWHP | PWDA2 | PWDA1 | PWAD | SMDA2 | SMDA1 | SMAD | RSTN |
| 01H | Analog Mute | 0 | 0 | 0 | 0 | AMT2RNAMT2LN | AMT1R | AMT1L |  |
| 02H | Interface Control | CKSB1 | CKSB0 | DIFC1 | DIFC0 | DIFB | 0 | DIFA1 | DIFA0 |
| 03H | DAC Speed Control | 0 | ACKS2 | DFS21 | DFS20 | 0 | ACKS1 | DFS11 | DFS10 |
| 04H | De-emphasis/ ATT speed | DEM21 | DEM20 | DEM11 | DEM10 | 0 | 0 | ATSAD | ATSDA |
| 05H | ADC Lch Volume | IATL7 | IATL6 | IATL5 | IATL4 | IATL3 | IATL2 | IATL1 | IATL0 |
| 06H | ADC Rch Volume | IATR7 | IATR6 | IATR5 | IATR4 | IATR3 | IATR2 | IATR1 | IATR0 |
| 07H | DAC1 Lch Volume | OAT1L7 | OAT1L6 | OAT1L5 | OAT1L44 | OAT1L3 | OAT1L2 | OAT1L1 | OAT1L0 |
| 08H | DAC1 Rch Volume | OAT1R7 | OAT1R6 | OAT1R5 | OAT1R4 | OAT1R3 | OAT1R2 | OAT1R1 | OAT1R0 |
| 09H | DAC2 Lch Volume | OAT2L7 | OAT2L6 | OAT2L5 | OAT2L4 | OAT2L3 | OAT2L2 | OAT2L1 | OAT2L0 |
| 0AH | DAC2 Rch Volume | OAT2R7 | OAT2R6 | OAT2R5 | OAT2R4 | OAT2R3 | OAT2R2 | OAT2R1 | OAT2R0 |
| 0BH | Headphone Control 1 | PTSA | 0 | HPZ | MOFF | HPMTN | PTS2 | PTS1 | PTS0 |
| 0CH | Headphone Control 2 | 0 | 0 | 0 | HPGA4 | HPGA3 | HPGA2 | HPGA1 | HPGA0 |
| 10H | Headphone Control 3 | 0 | 0 | AMTS2 | AMTS1 | AMTS0 | 0 | 0 | 0 |

Note: Data must not be written to the addresses from 0DH to 1FH. (except 10H)
When the PDN pin = "L", the registers are initialized to their default values.
When RSTN bit = " 0 ", the internal timing is reset, but registers are not initialized to their default values.
The bits defined as 0 must contain a " 0 " value.

## ■ Register Definitions

| Addr | Register Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 00H | Powerdown | PWHP | PWDA2 | PWDA1 | PWAD | SMDA2 | SMDA1 | SMAD | RSTN |  |  |  |  |  |  |  |  |  |
| R/W |  |  |  |  |  |  |  |  |  |  | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Default | 0 | 1 | 1 | 1 | 0 | 0 | 0 | 1 |  |  |  |  |  |  |  |  |  |  |

RSTN: Internal timing reset
0: Reset. Control Registers are NOT initialized.
1: Normal operation (default)
SMAD: ADC Digital Soft Mute Enable
0 : Normal operation (default)
1: ADC outputs soft-muted
SMDA1: DAC1 Digital Soft Mute Enable
0 : Normal operation (default)
1: All DAC outputs soft-muted
SMDA2: DAC2 Digital Soft Mute Enable
0 : Normal operation (default)
1: All DAC outputs soft-muted
PWAD: Power-down control of ADC
0: Power-down
1: Normal operation (default)
PWDA1: Power-down control of DAC1
0: Power-down
1: Normal operation (default)
PWDA2: Power-down control of DAC2
0: Power-down
1: Normal operation (default)
PWHP: Power-down control of Headphone Amplifier
0 : Power-down (default)
1: Normal operation

| Addr | Register Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 01 H | Analog Mute | 0 | 0 | 0 | 0 | AMT2RN | AMT2LN | AMT1RN | AMT1LN |  |  |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |  | R/W | RD | RD | RD | RD | R/W | R/W | R/W | R/W |
|  | Default | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |  |  |  |  |  |  |  |  |  |  |

AMT1LN: Analog Soft Mute Control for DAC1 Lch.

| MT1LN <br> Pin | AMT1LN <br> bit | DAC1Lch Analog Mute Status |
| :---: | :---: | :---: |
| L | 0 | Mute |
| L | 1 | Mute |
| H | 0 | Mute |
| H | 1 | Unmute |

Table 23. DAC1 Lch Analog Mute Control
AMT1RN: Analog Soft Mute Control for DAC1 Rch.

| MT1RN <br> Pin | AMT1RN <br> bit | DAC1Rch Analog Mute Status |
| :---: | :---: | :---: |
| L | 0 | Mute |
| L | 1 | Mute |
| H | 0 | Mute |
| H | 1 | Unmute |
| (default) |  |  |

Table 24. DAC1 Rch Analog Mute Control
AMT2LN, AMT2RN: Analog Soft Mute Control for DAC2.

| MT2LN <br> Pin | AMT2LN <br> bit | AMT2RN <br> bit | DAC2 Analog Mute Status |
| :---: | :---: | :---: | :---: |
| L | 0 | 0 | Mute |
| L | 0 | 1 | N/A |
| L | 1 | 0 | N/A |
| L | 1 | 1 | Mute |
| H | 0 | 0 | Mute |
| H | 0 | 1 | N/A |
| H | 1 | 0 | N/A |
| H | 1 | 1 | Unmule |
| (default) |  |  |  |

Table 25. DAC2 Analog Mute Control

| Addr | Register Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 02 H | Interface Settings | CKSB1 | CKSB0 | DIFC1 | DIFC0 | DIFB | 0 | DIFA1 | DIFA0 |
|  | R/W | R/W | R/W | R/W | R/W | R/W | RD | R/W | R/W |
|  | Default | 0 | 0 | 1 | 1 | 1 | 0 | 1 | 1 |

DIFA1-0: Audio format control for PORTA
(Default: I2S)
DIFB: Audio format control for PORTB
(Default: I2S)
DIFC1-0: Audio format control for PORTC
(Default: I2S)
CKSB1-0: ADC Clock control for Master mode.
(Default: 256fs)

| Addr | Register Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 03 H | DAC Speed Control | 0 | ACKS2 | DFS21 | DFS20 | 0 | ACKS1 | DFS11 | DFS10 |
|  |  |  |  |  |  |  |  |  |  |
| R/W | RD | R/W | R/W | R/W | RD | R/W | R/W | R/W |  |
| Default | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |  |

DFS11-10: DAC1 Sampling Speed Control
(Default: Normal Speed Mode)
These settings are ignored in Auto Setting Mode.
ACKS1: DAC1 Auto Setting Mode
0: Disable, Manual Setting Mode (default)
1: Enable, Auto Setting Mode
When ACKS1 bit = " 1 ", the master clock frequency is detected automatically and the DFS11-10 bits are ignored. When ACKS1 bit $=$ "0", DFS11-10 bits set the sampling speed mode.

DFS21-20: DAC2 Sampling Speed Control
(Default: Normal Speed Mode)
These settings are ignored in Auto Setting Mode.
ACKS2: DAC2 Auto Setting Mode
0: Disable, Manual Setting Mode (default)
1: Enable, Auto Setting Mode
When ACKS2 bit = " 1 ", the master clock frequency is detected automatically and the DFS21-20 bits are ignored. When ACKS2 bit = "0", DFS21-20 bits set the sampling speed mode.

| Addr | Register Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 04 H | De-emphasis/ ATT speed | DEM21 | DEM20 | DEM11 | DEM10 | 0 | 0 | ATSAD | ATSDA |
| R/W | R/W | R/W | R/W | R/W | RD | RD | R/W | R/W |  |
|  | Default | 0 | 1 | 0 | 1 | 0 | 0 | 0 | 0 |

ATSDA: DAC1/2 digital Attenuator transition time control
ATSAD: ADC digital Attenuator transition time control
DEM11-10: DAC1 De-emphasis filter control
DEM21-20: DAC2 De-emphasis filter control

| Addr | Register Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 05 H | ADC Lch Volume | IATL7 | IATL6 | IATL5 | IATL4 | IATL3 | IATL2 | IATL1 | IATL0 |
| 06 H | ADC Rch Volume | IATR7 | IATR6 | IATR5 | IATR4 | IATR3 | IATR2 | IATR1 | IATR0 |
|  |  |  |  |  |  |  |  |  |  |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |  |
|  | Default | 0 | 0 | 1 | 1 | 0 | 0 | 0 | 0 |

IATL7-0, IATR7-0: ADC Volume level control
(Default: 0dB)

| Addr | Register Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 07H | DAC1 Lch Volume | OAT1L7 | OAT1L6 | OAT1L5 | OAT1L4 | OAT1L3 | OAT1L2 | OAT1L1 | OAT1L0 |
| 08H | DAC1 Rch Volume | OAT1R7 | OAT1R6 | OAT1R5 | OAT1R4 | OAT1R3 | OAT1R2 | OAT1R1 | OAT1R0 |
| 09H | DAC2 Lch Volume | OAT2L7 | OAT2L6 | OAT2L5 | OAT2L4 | OAT2L3 | OAT2L2 | OAT2L1 | OAT2L0 |
| 0AH | DAC2 Rch Volume | OAT2R7 | OAT2R6 | OAT2R5 | OAT2R4 | OAT2R3 | OAT2R2 | OAT2R1 | OAT2R0 |
|  | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  | Default | 0 | 0 | 0 | 1 | 1 | 0 | 0 | 0 |

OAT1L7-0, OAT1R7-0, OAT2L7-0, OAT2R7-0: DAC1/2 Volume level control (Default: 0dB)

| Addr | Register Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0BH | Headphone Control 1 | PTSA | 0 | HPZ | MOFF | HPMTN | PTS2 | PTS1 | PTS0 |
| R/W | R/W | RD | R/W | R/W | R/W | R/W | R/W | R/W |  |
| Default | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 |  |

PTS2-0: Headphone-Amp Mute ON/OFF Transition Time
Default: "100"; typ. 16.4ms
HPMTN: Headphone-Amp Mute
0 : Mute (default)
1: Normal Output
MOFF: Soft transition for HPMTN bit change
0 : Enable (default)
1: Disable
HPZ: Headphone-Amp Pull-down Control
0 : Ground Mode (default)
HPL/HPR pins are shorted to VSS3.
1: Hi-Z Mode HPL/HPR pins are pulled-down by $50 \mathrm{k} \Omega(\mathrm{typ})$ to VSS5.

PTSA: MUTE pin/bit Transition Time Setting
0: Fixed (PTS2-0 = "011") (default)
1: Controlled by PTS2-0 bits

| Addr | Register Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 CH | Headphone Control 2 | 0 | 0 | 0 | HPGA4 | HPGA3 | HPGA2 | HPGA1 | HPGA0 |  |  |  |  |  |  |  |  |  |
| R/W |  |  |  |  |  |  |  |  |  |  | RD | RD | RD | R/W | R/W | R/W | R/W | R/W |
|  | Default | 0 | 0 | 0 | 1 | 1 | 0 | 0 | 1 |  |  |  |  |  |  |  |  |  |

HPGA4-0: Headphone-Amp Volume Setting
Default: 19H; 0dB
Refer Table 19.

| Addr | Register Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 10 H | Headphone Control 3 | 0 | 0 | AMTS2 | AMTS1 | AMTS0 | 0 | 0 | 0 |
| R/W | RD | RD | R/W | R/W | R/W | RD | RD | RD |  |
| Default | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |  |

AMTS2-0: Analog Mute Clock Source Control
Default: "000"; typ. 10.3ms
Refer Table 22.

## SYSTEM DESIGN

Figure 28 shows the system connection diagram. An evaluation board is available which demonstrates application circuits, the optimum layout, power supply arrangements and measurement results.


Figure 28. Typical Connection Diagram (Master Mode)
Notes:

- VSS1-7 must be connected the same analog ground plane.


## 1. Grounding and Power Supply Decoupling

The AK4685 requires careful attention to power supply and grounding arrangements. AVDD1, AVDD2, AVDD3, DVDD1, DVDD2, DVDD3, TVDD and PVDD are usually supplied from analog supply in system. VSS1-7 of the AK4685 must be connected to analog ground plane. System analog ground and digital ground must be connected separately near to where the supplies are brought onto the printed circuit board. Decoupling capacitors should be as near to the AK4685 as possible, with the small value ceramic capacitor being the nearest.

## 2. Voltage Reference Inputs

The voltage of AVDD1 sets the ADC input range, AVDD2(AVDD3) sets the $\mathrm{DAC1}(\mathrm{DAC} 2)$ analog output range. Normally, $0.1 \mu \mathrm{~F}$ ceramic capacitors should be connected between AVDD1/2/3 pins and VSS $6 / 2 / 3$ pins. The VCOM pin is a signal ground of this chip. An electrolytic capacitor $10 \mu \mathrm{~F}$ parallel with a $0.1 \mu \mathrm{~F}$ ceramic capacitor attached between these VCOM pins and VSS6 pin eliminates the effects of high frequency noise. No load current may be drawn from these VCOM pins. All signals, especially clocks, should be kept away from the AVDD1, AVDD2, AVDD3, and VCOM pins in order to avoid unwanted coupling into the AK4685.

## 3. Analog Inputs

The AK4685 receives the analog input through the single-ended Pre-amp using external resistors. The input range is $+/-3.3 \times$ AVDD $1 / 5 \mathrm{Vpp}$ (typ. $\mathrm{fs}=48 \mathrm{kHz}$ ) at each analog input pins. Each input pins are biased to $0 \mathrm{~V}(\mathrm{typ})$ internally. The ADC output data format is 2 's complement. The internal digital HPF removes the DC offset.

The AK4685 samples the analog inputs at 64fs. The digital filter rejects noise above the stop band except for multiples of 64 fs . The AK4685 includes an anti-aliasing filter (RC filter) to attenuate a noise around 64 fs .

## 4. Analog Outputs

The DAC1 outputs can be switched between single-ended and differential outputs.
When differential output is selected, the output range is $+/-2.56 \times(\operatorname{AVDD} 2) / 5 \mathrm{Vpp}(\mathrm{typ})$. The input data format is two's complement. The output voltage is positive full scale for 7FFFFFH (@24-bit) and negative full scale for 800000 H (@24-bit). The ideal voltage is 0 V for 000000 H (@24-bit). The internal switched-capacitor filter (SCF) attenuates the noise generated by the delta-sigma modulator beyond the audio passband.
When single-ended output is selected, the output range is $+/-1.41 \times(\mathrm{AVDD} 2) / 5 \mathrm{Vpp}(\mathrm{typ})$ centered around the VCOM voltage. The internal switched-capacitor filter (SCF) and continuous-time filter (CTF) attenuate the noise generated by the delta-sigma modulator beyond the audio passband.

The DAC2 outputs are single-ended output and it is for headphones. The output range is $+/-1.71 \times(A V D D 3) / 5 \mathrm{Vpp}(\mathrm{typ})$ centered around the 0 V . The input data format is two's complement. The output voltage is positive full scale for 7FFFFFH (@24-bit) and negative full scale for 800000 H (@24-bit). The ideal voltage is 0 V for 000000 H (@24-bit). The internal switched-capacitor filter (SCF) and continuous-time filter (CTF) attenuate the noise generated by the delta-sigma modulator beyond the audio passband.

DC offsets on the analog outputs should be eliminated by AC coupling since the analog outputs have a DC offset.

## 5. Attention to the PCB Wiring

Analog input and output pins should be wired as short as possible in order to avoid unwanted coupling into the AK4685.


## ■ Material \& Lead finish

Package molding compound: Epoxy, Halogen (bromine and chlorine) free Lead frame material: Cu
Lead frame surface treatment: Solder ( Pb free) plate


1) Pin \#1 indication
2) Asahi Kasei Logo
3) Marking Code: AK4685EQ
4) Date Code: XXXXXXX (7 digits)

## REVISION HISTORY

| Date (YY/MM/DD) | Revision | Reason | Page | Contents |
| :--- | :--- | :--- | :--- | :--- |
| $09 / 08 / 18$ | 00 | First Edition |  |  |

## IMPORTANT NOTICE

- These products and their specifications are subject to change without notice.

When you consider any use or application of these products, please make inquiries the sales office of Asahi Kasei Microdevices Corporation (AKM) or authorized distributors as to current status of the products.

- AKM assumes no liability for infringement of any patent, intellectual property, or other rights in the application or use of any information contained herein.
- Any export of these products, or devices or systems containing them, may require an export license or other official approval under the law and regulations of the country of export pertaining to customs and tariffs, currency exchange, or strategic materials.
- AKM products are neither intended nor authorized for use as critical components ${ }_{\text {Notel) }}$ in any safety, life support, or other hazard related device or system ${ }_{\text {Note2) }}$, and AKM assumes no responsibility for such use, except for the use approved with the express written consent by Representative Director of AKM. As used here:

Note1) A critical component is one whose failure to function or perform may reasonably be expected to result, whether directly or indirectly, in the loss of the safety or effectiveness of the device or system containing it, and which must therefore meet very high standards of performance and reliability.
Note2) A hazard related device or system is one designed or intended for life support or maintenance of safety or for applications in medicine, aerospace, nuclear energy, or other fields, in which its failure to function or perform may reasonably be expected to result in loss of life or in significant injury or damage to person or property.

- It is the responsibility of the buyer or distributor of AKM products, who distributes, disposes of, or otherwise places the product with a third party, to notify such third party in advance of the above content and conditions, and the buyer or distributor agrees to assume any and all responsibility and liability for and hold AKM harmless from any and all claims arising from the use of said product in the absence of such notification.

